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Title:

APPARATUS AND METHOD FOR ENCODING/DECODING TRANSPORT FORMAT COMBINATION INDICATOR IN CDMA MOBILE COMMUNICATION SYSTEM;

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ABSTRACT:

An apparatus and method for encoding/decoding a transport format combination indicator (TFCI) in a CDMA mobile communication system. In the TFCI encoding apparatus, a one-bit generator generates a sequence having the same symbols. A basis orthogonal sequence generator generates a plurality of basis orthogonal sequences. A basis mask sequence generator generates a plurality of basis mask sequences. An operation unit receives TFCI bits that are divided into a first information part representing biorthogonal sequence conversion, a second information part representing orthogonal sequence conversion, and a third information part representing mask sequence conversion and combines an orthogonal sequence selected from the basis orthogonal sequence based on the second information, a biorthogonal sequence obtained by combining the selected orthogonal sequence with the same symbols selected based on the first information part, and a mask sequence selected based on the biorthogonal sequence and the third information part, thereby generating a TFCI sequence.

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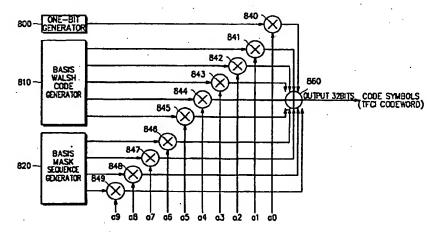
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(54) Title: APPARATUS AND METHOD FOR ENCODING/DECODING TRANSPORT FORMAT COMBINATION INDICATOR IN CDMA MOBILE COMMUNICATION SYSTEM



(57) Abstract: An apparatus and method for encoding/decoding a transport format combination indicator (TFCI) in a CDMA mobile communication system. In the TFCI encoding apparatus, a one-bit generator generates a sequence having the same symbols. A basis orthogonal sequence generator generates a plurality of basis orthogonal sequences. A basis mask sequence generator generates a plurality of basis mask sequences. An operation unit receives TFCI bits that are divided into a first information part representing biorthogonal sequence conversion, a second information part representing orthogonal sequence conversion, and a third information part representing mask sequence conversion and combines an orthogonal sequence selected from the basis orthogonal sequence based on the second information, a biorthogonal sequence obtained by combining the selected orthogonal sequence with the same symbols selected based on the first information part, and a mask sequence selected based on the biorthogonal sequence and the third information part, thereby generating a TFCI sequence.



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APPARATUS AND METHOD FOR ENCODING/DECODING TRANSPORT FORMAT COMBINATION INDICATOR IN CDMA MOBILE COMMUNICATION SYSTEM

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates generally to an information transmitting apparatus and method in an IMT 2000 system, and in particular, to an apparatus and method for transmitting a transport format combination indicator (TFCI).

2. Description of the Related Art

A CDMA mobile communication system (hereinafter, referred to as an IMT 2000 system) generally transmits frames that provide a voice service, an image service, a character service on a physical channel such as a dedicated physical data channel (DPDCH) at a fixed or variable data rate. In the case where the data frames which include that sort of services are transmitted at a fixed data rate, there is no need to inform a receiver of the spreading rate of each data frame. On the other hand, if the data frames are transmitted at a variable data rate, which implies that each data frame has a different data rate, a transmitter should inform the receiver of the spreading rate of each data frame determined by its data rate. A data rate is proportional to a data transmission rate and the data transmission rate is inversely proportional to a spreading rate in a general IMT 2000 system.

For transmission of data frames at a variable data rate, a TFCI field of a DPCCH informs a receiver of the data rate of the current service frame. The TFCI field includes a TFCI indicating a lot of information including the data rate of a service frame. The TFCI is information that helps a voice or data service to reliably be provided.

FIGs. 1A to 1D illustrate examples of applications of a TFCI. FIG. 1A illustrates application of the TFCI to an uplink DPDCH and an uplink dedicated physical control channel (DPCCH). FIG. 1B illustrates application of the TFCI to a random access channel (RACH). FIG. 1C illustrates application of the TFCI to a downlink DPDCH and a downlink DPCCH. FIG. 1D illustrates application of the TFCI to a secondary common control physical channel (SCCPCH).

Referring to FIGs. 1A to 1D, one frame is comprised of 16 slots and each slot has a TFCI field. Thus, one frame includes 16 TFCI fields. A TFCI field includes N_{TFCI} bits and a TFCI generally has 32 bits in a frame. To transmit the 32-bit TFCI in one frame, 2 TFCI bits can be assigned to each of the 16 slots ($T_{slot} = 0.625$ ms).

FIG. 2 is a block diagram of a base station transmitter in a general IMT 2000 system.

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Referring to FIG. 2, multipliers 211, 231, and 232 multiply input signals by gain coefficients G₁, G₃, and G₅. Multipliers 221, 241, and 242 multiply TFCI codewords (TFCI code symbols) received from corresponding TFCI encoders by gain coefficients G₂, G₄, and G₆. The gain coefficients G₁ to G₆ may have different values according to service types or handover situations. The input signals include pilots and power control signals (TPCs) of a DPCCH and a DPDCH data. A multiplexer 212 inserts 32 bit TFCIcode symbols(TFCI codeword) received from the multiplier 221 into the TFCI fields as shown in FIG 1C. A multiplexer 242 inserts 32 bit TFCI code symbols received from the multiplier 241 into the TFCI fields. A multiplexer 252 inserts 32 bit TFCI code symbols received from the multiplier 242 into the TFCI fields. Insertion of TFCI code symbols into TFCI fields is shown in FIGs. 1A to 1D. The 32 code symbols are obtained by encoding TFCI bits(information bits) that define the data rate of a data signal on a corresponding data channel. 1st, 2nd, and 3rd serial to parallel converters (S/Ps) 213, 233, and 234 separate the outputs of the multiplexers 212, 242, and 252 into I channels and Q channels. Multipliers 214, 222, and 235 to 238 multiply the outputs of the S/Ps 213, 233, and 234 by channelization codes C_{ch1}, C_{ch2}, and C_{ch3}. The channelization codes are orthogonal codes. A first summer 215 sums the outputs of the multipliers 214, 235, and 237 and generates an I channel signal and a second summer 223 sums the outputs of the multipliers 222, 236, and 238 and generates a Q channel signal. A phase shifter 224 shifts the phase of the Q channel signal received from the second summer 223 by 90°. A summer 216 adds the outputs of the first summer 215 and the phase shifter 224 and generates a complex signal I+jQ. A multiplier 217 scrambles the complex signal with a complex PN sequence C_{scramb} assigned to the base station. A signal processor(S/P) 218 separates the scrambled signal into an I channel and a Q channel. Low-pass filters (LPFs) 219 and 225 limits the bandwidths of the I channel and Q channel signals received from the S/P 218 by low-pass-filtering. Multipliers 220 and 226 multiply the outputs of the LPFs 219 and 225 by carriers $\cos(2\pi f_c t)$ and $\sin(2\pi f_c t)$, respectively, thereby transforming the outputs of the LPFs 219 and 225 to an RF (Radio

Frequency) band. A summer 227 sums the RF I channel and Q channel signals.

FIG. 3 is a block diagram of a mobile station transmitter in the general IMT 2000 system.

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Referring to FIG. 3, multipliers 311, 321, and 323 multiply corresponding signals by channelization codes C_{ch1}, C_{ch2}, and C_{ch3}. Signals 1, 2, 3 are first, second and third DPDCH signal. An input signal 4 includes pilots and TPCs of a DPCCH.TFCI information bits are encoded into 32 bit TFCI code symbols by a TFCI encoder 309. A multiplier 310 inserts a 32 bit TFCI code symbols into the signal 4 as shown in FIG. 1A. A multiplier 325 multiplies multiplies a DPCCH signal which include TFCI code symbol received from the multiplier 310 by a channelization code C_{che}. channelization codes C_{ch1} to C_{ch4} are orthogonal codes. The 32 TFCI code symbols are obtained by encoding TFCI information bits that define the data rate of the DPDCH signals. Multipliers 312, 322, 324, and 326 multiply the outputs of the multipliers 311, 321, 323, and 325 by gain coefficients G_1 to G_4 , respectably. The gain coefficients G_1 to G₄ may have different values. A first summer 313 generates an I channel signal by adding the outputs of the multipliers 312 and 322. A second summer 327 generates a Q channel signal by adding the outputs of the multipliers 324 and 326. A phase shifter 328 shifts the phase of the Q channel signal received from the second summer 327 by 90°. A summer 314 adds the outputs of the first summer 313 and the phase shifter 328 and generates a complex signal I+jQ. A multiplier 315 scrambles the complex signal with a PN sequence C_{scramb} assigned to a base station. An S/P 329 divides the scrambled signal into an I channel and a Q channel. LPFs 316 and 330 low-pass-filter the I channel and Q channel signals received from the S/P 329 and generate signals with limited bandwidths. Multipliers 317 and 331 multiply the outputs of the LPFs 316 and 330 by carriers $\cos(2\pi f.t)$ and $\sin(2\pi f.t)$, respectively, thereby transforming the outputs of the LPFs 316 and 330 to an RF band. A summer 318 sums the RF I channel and Q channel signals.

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TFCIs are categorized into a basic TFCI and an extended TFCI. The basic TFCI represents 1 to 64 different information including the data rates of corresponding data channels using 6 TFCI information bits, whereas the extended TFCI represents 1 to 128, 1 to 256, 1 to 512, or 1 to 1024 different information using 7, 8, 9 or 10 TFCI information bits. The extended TFCI has been suggested to satisfy the requirement of the IMT 2000 system for more various services. TFCI bits are essential for a receiver to receive data frames received from a transmitter. That is the reason why unreliable transmission of the TFCI information bits due to transmission errors lead to wrong

interpretation of the frames in the receiver. Therefore, the transmitter encodes the TFCI bits with an error correcting code prior to transmission so that the receiver can correct possibly generated errors in the TFCI.

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FIG. 4A conceptionally illustrates a basic TFCI bits encoding structure in a conventional IMT 2000 system and FIG. 4B is an exemplary encoding table applied to a biorthogonal encoder shown in FIG. 4A. As stated above, the basic TFCI has 6 TFCI bits (hereinafter, referred to as basic TFCI bits) that indicate 1 to 64 different information.

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Referring to FIGs. 4A and 4B, a biorthogonal encoder 402 receives basic TFCI bits and outputs 32 coded symbols (TFCI codeword or TFCI code symbol). The basic TFCI is basically expressed in 6 bits. Therefore, in the case where a basic TFCI bits of less than 6 bits are applied to the biorthogonal encoder 402, 0s are added to the left end, i.e., MSB (Most Significant Bit) of the basic TFCI bits to increase the number of the basic TFCI bits to 6. The biorthogonal encoder 402 has a predetermined encoding table as shown in FIG. 4B to output 32 coded symbols for the input of the 6 basic TFCI bits. As shown in FIG. 4B, the encoding table lists 32(32-symbol) orthogonal codewords $c_{32.1}$ to $c_{32.32}$ and 32 biorthogonal codewords $\overline{c_{32.1}}$ to $\overline{c_{32.32}}$ that are the complements of the codewords $c_{32.1}$ to $c_{32.32}$. If the LSB (Least Significant Bit) of the basic TFCI is 1, the biorthogonal encoder 402 selects out of the 32 biorthogonal codewords. If the LSB is 0, the biorthogonal encoder 402 selects out of the 32 orthogonal codewords. One of the selected orthogonal codewords or biorthogonal codewords is then selected based on the other TFCI bits.

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A TFCI codeword should have powerful error correction capability as stated before. The error correction capability of binary linear codes depends on the minimum distance (dmin) between the binary linear codes. A minimum distance for optimal binary linear codes is described in "An Updated Table of Minimum-Distance Bounds for Binary Linear Codes", A.E. Brouwer and Tom Verhoeff, IEEE Transactions on Information Theory, vol. 39, No. 2, March 1993 (hereinafter, referred to as reference 1).

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Reference 1 gives 16 as a minimum distance for binary linear codes by which 32 bits are output for the input of 6 bits. TFCI codewords output from the biorthogonal encoder 402 has a minimum distance of 16, which implies that the TFCI codewords are optimal codes.

FIG. 5A conceptionally illustrates an extended TFCI bits encoding structure in the conventional IMT 2000 system, FIG. 5B is an exemplary algorithm of distributing TFCI bits in a controller shown in FIG. 5A, and FIG. 5C illustrates an exemplary encoding table applied to biorthogonal encoders shown in FIG. 5A. An extended TFCI is also defined by the number of TFCI bits. That is, the extended TFCI includes 7, 8, 9 or 10 TFCI bits (hereinafter, referred to as extended TFCI bits) that represent 1 to 128, 1 to 256, 1 to 512, or 1 to 1024 different information, as stated before.

Referring to FIGs. 5A, 5B, and 5C, a controller 500 divides TFCI bits into two halves. For example, for the input of 10 extended TFCI bits, the controller 500 outputs the first half of the extended TFCI as first TFCI bits (word 1) and the last half as second TFCI bits (word 2). The extended TFCI are basically expressed in 10 bits. Therefore, in the case where an extended TFCI bits of less than 10 bits are input, the controller 500 adds 0s to the MSB of the extended TFCI bits to represent the extended TFCI in 10 bits. Then, the controller 500 divides the 10 extended TFCI bits into word 1 and word 2. Word 1 and word 2 are fed to biorthogonal encoders 502 and 504, respectively. A method of separating the extended TFCI bits a_1 to a_{10} into word 1 and word 2 is illustrated in FIG. 5B.

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The biorthogonal encoder 502 generates a first TFCI codeword having 16 symbols by encoding word 1 received from the controller 500. The biorthogonal encoder 504 generates a second TFCI codeword having 16 symbols by encoding word 2 received from the controller 500. The biorthogonal encoders 502 and 504 have predetermined encoding tables to output the 16-symbol TFCI codewords for the two 5-bit TFCI inputs (word 1 and word 2). An exemplary encoding table is illustrated in FIG. 5C. As shown in FIG. 5C, the encoding table lists 16 orthogonal codewords of length 16 bits $c_{16.1}$ to $c_{16.16}$ and biorthogonal codewords $c_{16.1}$ to $c_{16.16}$ that are the complements of the 16 orthogonal codewords. If the LSB of 5 TFCI bits is 1, a biorthogonal encoder (502 or 504) selects the 16 biorthogonal codewords. If the LSB is 0, the biorthogonal encoder selects the 16 orthogonal codewords. Then, the biorthogonal encoder selects one of the selected orthogonal codewords or biorthogonal codewords based on the other TFCI bits and outputs the selected codeword as the first or second TFCI codeword.

A multiplexer 510 multiplexes the first and second TFCI codewords to a final 32-symbol TFCI codeword.

Upon receipt of the 32-symbol TFCI codeword, a receiver decodes the TFCI codeword separately in halves (word 1 and word 2) and obtains 10 TFCI bits by combining the two decoded 5-bit TFCI halves. In this situation, a possible error even in one of the decoded 5-bit TFCI output during decoding leads to an error over the 10 TFCI bits.

An extended TFCI codeword also should have a powerful error correction capability. To do so, the extended TFCI codeword should have the minimum distance as suggested in reference 1.

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In consideration of the number 10 of extended TFCI bits and the number 32 of the symbols of a TFCI codeword, reference 1 gives 12 as a minimum distance for an optimal code. Yet, a TFCI codeword output from the structure shown in FIG. 5A has a minimum distance of 8 because an error in at least one of word 1 and word 2 during decoding results in an error in the whole 10 TFCI bits. That is, although extended TFCI bits are encoded separately in halves, a minimum distance between final TFCI codewords is equal to a minimum distance 8 between codeword outputs of the biorthogonal encoders 502 and 504.

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Therefore, a TFCI codeword transmitted from the encoding structure shown in FIG. 5A is not optimal, which may increase an error probability of TFCI bits in the same radio channel environment. With the increase of the TFCI bit error probability, the receiver misjudges the data rate of received data frames and decodes the data frames with an increased error rate, thereby decreasing the efficiency of the IMT 2000 system.

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According to the conventional technology, separate hardware structures are required to support the basic TFCI and the extended TFCI. As a result, constraints are imposed on implementation of an IMT 2000 system in terms of cost and system size.

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SUMMARY OF THE INVENTION

It is, therefore, an object of the present invention to provide an apparatus and method for encoding an extended TFCI in an IMT 2000 system.

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It is also an object of the present invention to provide an apparatus and method for encoding a basic TFCI and an extended TFCI compatibly in an IMT 2000 system.

It is another object of the present invention to provide an apparatus and method for decoding an extended TFCI in an IMT 2000 system.

It is still another object of the present invention to provide an apparatus and method for decoding a basic TFCI and an extended TFCI compatibly in an IMT 2000 system.

It is yet another object of the present invention to provide an apparatus and method for generating an optimal code by encoding an extended TFCI in an IMT 2000 system.

It is a further object of the present invention to provide a method of generating mask sequences for use in encoding/decoding an extended TFCI in an IMT 2000 system.

To achieve the above objects, there is provided a TFCI encoding/decoding apparatus and method in a CDMA mobile communication system. In the TFCI encoding apparatus, a one-bit generator generates a sequence having the same symbols. A basis orthogonal sequence generator generates a plurality of basis orthogonal sequences. A basis mask sequence generator generates a plurality of basis mask sequences. An operation unit receives TFCI bits that are divided into a 1st information part representing biorthogonal sequence conversion, a 2nd information part representing orthogonal sequence conversion and combines an orthogonal sequence selected from the basis orthogonal sequence based on the 2nd information, a biorthogonal sequence obtained by combining the selected orthogonal sequence with the same symbols selected based on the 1st information part, and a mask sequence selected based on the biorthogonal code sequence and the 3rd information part, thereby generating a TFCI sequence.

BRIEF DESCRIPTION OF THE DRAWINGS

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The above and other objects, features and advantages of the present invention will become more apparent from the following detailed description when taken in conjunction with the accompanying drawings in which:

FIGs. 1A to 1D illustrate exemplary applications of a TFCI to channel frames in a general IMT 2000 system;

FIG. 2 is a block diagram of a base station transmitter in the general IMT 2000 system;

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- FIG. 3 is a block diagram of a mobile station transmitter in the general IMT 2000 system;
- FIG. 4A conceptionally illustrates a basic TFCI encoding structure in a conventional IMT 2000 system;
- FIG. 4B is an example of an encoding table used in a biorthogonal encoder shown in FIG. 4A;
- FIG. 5A conceptionally illustrates an extended TFCI encoding structure in the conventional IMT 2000 system;
- FIG. 5B is an example of an algorithm of distributing TFCI bits in a controller shown in FIG. 5A;
- FIG. 5C is an example of an encoding table used in biorthogonal encoders shown in FIG. 5A;
- FIG. 6 conceptionally illustrates a TFCI encoding structure in an IMT 2000 system according to the present invention;
- FIG. 7 is a flowchart illustrating an embodiment of a mask sequence generating procedure for TFCI encoding in the IMT 2000 system according to the present invention;
- FIG. 8 is a block diagram of an embodiment of a TFCI encoding apparatus in the IMT 2000 system according to the present invention;
- FIG. 9 is a block diagram of an embodiment of a TFCI decoding apparatus in the IMT 2000 system according to the present invention;
- FIG. 10 is a flowchart illustrating a control operation of a correlation comparator shown in FIG. 9;
- FIG. 11 is a flowchart illustrating an embodiment of a TFCI encoding procedure in the IMT 2000 system according to the present invention;
- FIG. 12 is a flowchart illustrating another embodiment of the TFCI encoding procedure in the IMT 2000 system according to the present invention;
- FIG. 13 illustrates an embodiment of the structures of orthogonal sequences and mask sequences determined by a TFCI according to the present invention;
- FIG. 14 is a block diagram of another embodiment of the TFCI encoding apparatus in the IMT 2000 system according to the present invention;
- FIG. 15 is a block diagram of another embodiment of the TFCI decoding apparatus in the IMT 2000 system according to the present invention;
- FIG. 16 is a flowchart illustrating another embodiment of the TFCI encoding procedure in the IMT 2000 system according to the present invention; and
- FIG. 17 is a block diagram of a third embodiment of the TFCI decoding apparatus in the IMT 2000 system according to the present invention.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Preferred embodiments of the present invention will be described herein below with reference to the accompanying drawings. In the following description, well-known functions or constructions are not described in detail since they would obscure the invention in unnecessary detail.

The present invention is directed to a TFCI encoding concept of outputting final code symbols (a TFCI codeword) by adding first code symbols (a first TFCI codeword) resulting from first TFCI bits and second code symbols (a second TFCI codeword) resulting from second TFCI bits in an IMT 2000 system. The TFCI encoding concept is shown in FIG. 6. Here, a biorthogonal sequence and a mask sequence are given as the first TFCI codeword and the second TFCI codeword, respectively.

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Referring to FIG.6, TFCI bits are separated into the first TFCI bits and the second TFCI bits. A mask sequence generator 602 generates a predetermined mask sequence by encoding the second TFCI bits and a biorthogonal sequence generator 604 generates a predetermined biorthogonal sequence by encoding the first TFCI bits. An adder 610 adds the mask sequence and the biorthogonal sequence and outputs final code symbols (a TFCI codeword). The mask sequence generator 602 may have an encoding table that lists mask sequences for all possible second TFCI bits. The biorthogonal sequence generator 604 may also have an encoding table that lists biorthogonal sequences for all possible first TFCI bits.

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As described above, mask sequences and a mask sequence generating method should be defined to implement the present invention. Walsh codes are given as orthogonal sequences by way of example in embodiments of the present invention.

1. Mask Sequence Generating Method

The present invention pertains to encoding and decoding of TFCI bits and use of an extended Reed Muller code in an IMT 2000 system. For this purpose, predetermined sequences are used and the sequences should have a minimum distance that ensures excellent error correction performance.

A significant parameter that determines the performance or capability of a linear

error correcting code is a minimum distance between codewords of the error correcting code. The Hamming weight of a codeword is the number of its symbols other than 0. If a codeword is given as "0111", its Hamming weight is 3. The smallest Hamming weight of a codeword except all "0" codeword is called a minimum weight and the minimum distance of each binary linear code is equal to the minimum weight. A linear error correcting code has a better error correcting performance as its minimum distance is increased. For details, see "The Theory of Error-Correcting Codes", F.J. Macwilliams and N.J.A. Sloane, North-Holland (hereinafter, referred to as reference 2).

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An extended Reed Muller code can be derived from a set of sequences each being the sum of the elements of an m-sequence and a predetermined sequence. To use the sequence set as a linear error correcting code, the sequence set should have a large minimum distance. Such sequence sets include a Kasami sequence set, a Gold sequence set, and a Kerdock sequence set. If the total length of a sequence in such a sequence set is $L=2^{2m}$, a minimum distance = $(2^{2m}-2^m)/2$. For $L=2^{2m+1}$, the minimum distance = $(2^{2m+1}-2^{2m})/2$. That is, if L=32, the minimum distance = 12.

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A description will be made of a method of generating a linear error correcting code with excellent performance, i.e., an extended error correcting code (Walsh codes and mask sequences).

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According to a coding theory, there is a column transposition function for making Walsh codes from m-sequences in a group which has been formed by cyclically shifting an originating m-sequence by one to 'n' times, where the 'n' is a length of the m-sequence. In other words, each of the m-sequences is formed by cyclically shifting the originating m-sequence by a particular number of times. The column transposition function is a converting function which converts the swquences in the m-sequence group to Walsh codes. We assume there is a sequence such as a Gold sequence or a Kasami sequence which is formed by adding the originating m-sequence with another Another group of m-sequences is similarly formed by originating m-sequence. cyclically shifting the other originating m-sequence one to 'n' times, where 'n' is the length of the predetermined sequence. Afterwards, a reverse column transposition function is applied to the second group of m-sequences formed from the other originating m-sequence. The application of the reverse column transposition function to the second group of m-sequences creates another set of sequences which shall be defined as mask sequences.

In an embodiment of the present invention, a mask sequence generating method is described in connection with generation of a $(2^n, n+k)$ code (extended Reed Muller code) (here, k = 1, ..., n+1) using a Gold sequence set. The $(2^n, n+k)$ code represents output of a 2^n -symbol TFCI codeword for the input of (n+k) TFCI bits (input information bits). It is well known that a Gold sequence can be expressed as the sum of two different m-sequences. To generate the $(2^n, n+k)$ code, therefore, Gold sequences of length (2^n-1) should be produced. Here, a Gold sequence is the sum of two m-sequences $m_1(t)$ and $m_2(t)$ that are generated from generator polynomials fl(x) and f2(x). Given the generator polynomials fl(x) and f2(x), the m-sequences $m_1(t)$ and $m_2(t)$ are computed using a Trace function.

$$m_1(t) = Tr(A\alpha^t)$$
 $t = 0, 1, ..., 30$ and $Tr(a) = \sum_{k=0}^{n-1} \alpha^{2k}$, $a \in GF(2^n)$ (Eq. 1)

where A is determined by the initial value of an m-sequence, α is the root of the polynomial, and n is the order of the polynomial.

FIG. 7 is a flowchart illustrating a mask sequence generating procedure for use in generating a (2ⁿ, n+k) code from a Gold sequence set.

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Referring to FIG. 7, m-sequences $m_1(t)$ and $m_2(t)$ are generated in Eq. 1 using the generator polynomials fl(x) and f2(x), respectively in step 710. In step 712, a sequence transposition function $\sigma(t)$ is calculated to make Walsh codes from a sequence set having m-sequences formed by cyclically shifting $m_2(t)$ 0 to n-2 times where all '0' column is inserted in front of the m-sequences made from $m_2(t)$, as shown below:

$$\sigma: \{0, 1, 2, ..., 2^{n}-2\} \to \{1, 2, 3, ..., 2^{n}-1\}$$

$$\sigma(t) = \sum_{i=0}^{n-1} m_{2}(t+i)2^{n-1-i} \quad t = 0, 1, 2, ... \quad (Eq. 2)$$

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A set of 31 sequences produced by cyclically shifting the m-sequence $m_1(t)$ 0 to 30 times are column-transposed with the use of $\sigma^{-1}(t)+2$ derived from the reverse function of $\sigma(t)$ in step 730. Then, 0s are added to the start of each of the resulting column-transposed sequences to make the length of the sequence 2^n . Thus, a set $d_i(t)$ of (2^n-1) sequences of length 2^n ($i=0,...,2^n-2,t=1,...,2^n$) are generated.

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$$\{d_{i}(t) | t = 1, ..., 2^{n}, i = 0, ..., 2^{n} - 2\}$$

$$d_{i}(t) = \begin{pmatrix} 0, & if, t = 1 \\ m_{i}(\sigma^{-1}(t+i) + 2), & if, t = 2,3,...,2^{n} \end{pmatrix} \qquad (Eq. 3)$$

A plurality of $d_i(t)$ are mask functions that can be used as 31 masks.

 $d_i(t)$ is characterized in that two different masks among the above masks are added to one of (2^n-1) masks except for the two masks. To further generalize it, each of the (2^n-1) masks can be expressed as the sum of at least two of particular n masks. The n masks are called basis mask sequences. When the $(2^n, n+k)$ code is to be generated, the total number of necessary codewords is 2^{n+k} for n+k input information bits (TFCI bits). The number of 2^n orthogonal sequences (Walsh sequences) and their complements, i.e. biorthogonal sequences, is $2^n \times 2 = 2^{n+1}$. $2^{k-1}-1(=(2^{n+k}/2^{n+1})-1)$ masks that are not 0s are needed for generation of the $(2^n, n+k)$ code. Here, the $2^{k-1}-1$ masks can be expressed by the use of k-1 basis mask sequences, as stated before.

Now, a description will be given of a method of selecting the k-1 basis mask sequences. The m-sequence $m_1(t)$ is cyclically shifted 0 to 2^{n-1} times to generate a set of sequences in step 730 of FIG. 7. Here, an m-sequence obtained by cyclically shifting the m-sequence $m_1(t)$ i times is expressed as $Tr(\alpha^i, \alpha^i)$ according to Eq.1. That is, a set of sequences are generated by cyclically shifting the m-sequence $m_1(t)$ 0 to 30 times with respect to an initial sequence $A = \{1, \alpha, ..., \alpha^{2n-2}\}$. Here, linearly independent k-1 basis elements are found from the Galois elements 1, α , ..., $\alpha^{2^{n-2}}$ and mask sequences corresponding to the output sequences of a Trace function with the k-1 basis elements as an initial sequence become basis mask sequences. A linear independence condition is expressed as

$$\alpha_1, ..., \alpha_{k-1}$$
: linearly independent
$$\Leftrightarrow c_1\alpha_1 + c_2\alpha_2 + ... + c_{k-1}\alpha_{k-1} \neq 0, \quad \forall c_1, c_2, ..., c_{k-1} \qquad \qquad (Eq. 4)$$

To describe the above generalized mask function generation method in detail, how to generate a (32, 10) code using a Gold sequence set will be described referring to FIG. 7. It is well known that a Gold sequence is expressed as the sum of different predetermined m-sequences. Therefore, a Gold sequence of length 31 should be

generated first in order to generate the intended (32, 10) code. The Gold sequence is the sum of two m-sequences generated respectively from polynomials x^5+x^2+1 and x^5+x^4+x+1 . Given a corresponding generator polynomial, each of the m-sequences $m_1(t)$ and $m_2(t)$ is computed using a Trace function by

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$$m_1(t) = Tr(A\alpha^t)$$
 $t = 0, 1, ..., 30$ and $Tr(a) = \sum_{n=0}^{4} \alpha^{2^n}$, $a \in GF(2^5)$ (Eq. 5)

where A is determined by the initial value of the m-sequence, α is the root of the polynomial, and n is the order of the polynomial, here 5.

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FIG. 7 illustrates the mask function generating procedure to generate the (32, 10) code.

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Referring to FIG. 7, m-sequences $m_1(t)$ and $m_2(t)$ are generated in Eq. 1 using the generator polynomials fl(x) and f2(x), respectively in step 710. In step 712, the column transposition function $\sigma(t)$ is calculated to make a Walsh code of the m-sequence $m_2(t)$ by

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$$\sigma: \{0, 1, 2, ..., 30\} \to \{1, 2, 3, ..., 31\}$$

$$\sigma(t) = \sum_{i=0}^{4} m_2(t-i)2^{4-i} \qquad (Eq. 6)$$

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Then, a set of 31 sequences produced by cyclically shifting the m-sequence $m_i(t)$ 0 to 30 times are column-transposed with the use of $\sigma^{-1}(t)+2$ derived from the reverse function of $\sigma(t)$ in step 730. Then, 0s are added to the start of each of the resulting sequence-transposed sequences to make the length of the sequence 31. Thus, 31 $d_i(t)$ of length 32 are generated. Here, if i = 0, ..., 31, t = 1, ..., 32. The sequences set generated in step 730 can be expressed as

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$$\{d_{i}(t) | t = 1, ..., 32, i = 0, ..., 30\}$$

$$d_{i}(t) = \begin{pmatrix} 0, & if, t = 1 \\ m_{1}(\sigma^{-1}(t+i)+2), & if, t = 2,3,...,32 \end{pmatrix}(Eq. 7)$$

A plurality of d_i(t) obtained from Eq. 7 can be used as 31 mask sequences.

 $d_i(t)$ is characterized in that two different masks among the above masks are added to one of the 31 masks except for the two masks. In other words, each of the 31 masks can be expressed as a sum of 5 particular masks. These 5 masks are basis mask sequences.

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When the (32, 10) code is to be generated, the total number of necessary codewords is $2^n = 1024$ for all possible 10 input information bits (TFCI bits). The number of biorthogonal sequences of length 32 is 32 x 2 = 64. 15 masks are needed to generate the (32, 10) code. The 15 masks can be expressed as combinations of 4 basis mask sequences.

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Now, a description will be given of a method of selecting the 4 basis mask sequences. An m-sequence obtained by cyclically shifting the m-sequence $m_1(t)$ i times is expressed as $Tr(\alpha^i \cdot \alpha^i)$ according to Eq.1. That is, a set of sequences are generated by cyclically shifting the m-sequence $m_1(t)$ 0 to 30 times with respect to an initial sequence $A = \{1, \alpha, ..., \alpha^{2n-2}\}$. Here, 4 linearly independent basis elements are found from the Galois elements 1, α , ..., α^{2n-2} and mask sequences corresponding to the output sequences of a Trace function with the 4 basis elements as an initial sequence becoming basis mask sequences. A linear independence condition is expressed as

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$$\alpha$$
, β , γ , δ : linearly independent
 $\Leftrightarrow c_1\alpha + c_2\beta + c_3\gamma$, $+ c_4\delta \neq 0$, $\forall c_1, c_2, c_3, c_4$ (Eq. 8)

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In fact, 1, α , α^2 , α^3 in the Galois GF(2⁵) are polynomial sub-bases that are well known as four linearly independent elements. By replacing the variable A in Eq. 1 with the polynomial bases, four basis mask sequences M1, M2, M4, and M8 are achieved.

M1 = 001010000110001111111000001110111

M2 = 0000000111001101101101101111000111

M4 = 000010101111110010001101100101011

M8 = 00011100001101110010111101010001

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There will herein below be given a description of an apparatus and method for encoding/decoding a TFCI using basis mask sequences as obtained in the above manner in an IMT 2000 system according to embodiments of the present invention.

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2. First Embodiment of Encoding/Decoding Apparatus and Method

FIGs. 8 and 9 are block diagrams of TFCI encoding and decoding apparatuses in an IMT 2000 system according to an embodiment of the present invention.

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Referring to FIG. 8, 10 TFCI bits a0 to a9 are applied to corresponding multipliers 840 to 849. A one-bit generator 800 continuously generates a predetermined code bit. That is, since the present invention deals with biorthogonal sequences, necessary bits are generated to make a biorthogonal sequence out of an orthogonal sequence. For example, the one-bit generator 800 generates bits having 1s to inverse an orthogonal sequence (i.e., a Walsh code) generated from a basis Walsh code generator 810 and thus generate a biorthogonal sequence. The basis Walsh code generator 810 generates basis Walsh codes of a predetermined length. The basis Walsh codes refer to Walsh codes from which all intended Walsh codes can be produced through arbitrary addition. For example, when Walsh codes of length 32 are used, the basis Walsh codes are 1st, 2nd, 4th, 8th, and 16th Walsh codes W1, W2, W4, W8, and W16, wherein:

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W1:0101010101010101010101010101010101

W2:0011001100110011001100110011011

W4: 000011110000111100001111

W8: 0000000011111111100000000111111111

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A basis mask sequence generator 820 generates a basis mask sequence of a predetermined length. A basis mask sequence generating method has already been described before and its details will not be described. If a mask sequence of length 32 is used, basis mask sequences are 1st, 2nd, 4th, and 8th mask sequences M1, M2, M4, M8, wherein:

M1: 001010000110001111111000001110111

M2: 00000001110011010110110110111000111

M4: 000010101111110010001101100101011

M8: 00011100001101110010111101010001.

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The multiplier 840 multiplies 1s output from the one-bit generator 800 by the input information bit a0 on a symbol basis.

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The multiplier 841 multiplies the basis Walsh code W1 received from the basis Walsh code generator 810 by the input information bit a1. The multiplier 842 multiplies the basis Walsh code W2 received from the basis Walsh code generator 810 by the input

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information bit a2. The multiplier 843 multiplies the basis Walsh code W4 received from the basis Walsh code generator 810 by the input information bit a3. The multiplier 844 multiplies the basis Walsh code W8 received from the basis Walsh code generator 810 by the input information bit a4. The multiplier 845 multiplies the basis Walsh code W16 received from the basis Walsh code generator 810 by the input information bit a5. The multipliers 841 to 845 multiply the received basis Walsh codes W1, W2, W4, W8, and W16 by their corresponding input information bits symbol by symbol.

Meanwhile, the multiplier 846 multiplies the basis mask sequence M1 by the input information bit a6. The multiplier 847 multiplies the basis mask sequence M2 by the input information bit a7. The multiplier 848 multiplies the basis mask sequence M4 by the input information bit a8. The multiplier 849 multiplies the basis mask sequence M8 by the input information bit a9. The multipliers 846 to 849 multiply the received basis mask sequences M1, M2, M4, and M8 by their corresponding input information bits symbol by symbol.

An adder 860 adds the encoded input information bits received from the multipliers 840 to 849 and outputs final code symbols of length 32 bits (a TFCI codeword). The length of the final code symbols (TFCI codeword) is determined by the lengths of the basis Walsh codes generated from the basis Walsh code generator 810 and the basis mask sequences generated from the basis mask sequence generator 820.

For example, if the input information bits a0 to a9 are "0111011000", the multiplier 840 multiplies 0 as a0 by 1s received from the one-bit generator 800 and generates 32 code symbols being all "0s". The multiplier 841 multiplies 1 as a1 by W1 received from the basis Walsh code generator 810 and generates code symbols "01010101010101010101010101010101". The multiplier 842 multiplies 1 as a2 by W2 received from the basis Walsh code generator 810 and generates code symbols "00110011001100110011001100110011". The multiplier 843 multiplies 1 as a3 by W4 received from the basis Walsh code generator 810 and generates code symbols "00001111000011110000111100001111". The multiplier 844 multiplies 0 as a4 by W8 received from the basis Walsh code generator 810 and generates 32 code symbols being all "0s". The multiplier 845 multiplies 1 as a5 by W16 received from the basis Walsh multiplier 846 multiplies 1 as a6 by M1 received from the basis mask sequence generator 820 and generates "00101000011000111111000001110111". The multiplier 847 multiplies 0 as a7 by M2 received from the basis mask sequence generator 820 and generates 32 code symbols being all 0s. The multiplier 848 multiplies 0 as a8 by M4 received from the basis mask sequence generator 820 and generates 32 code symbols being all 0s. The multiplier 849 multiplies 0 as a9 by M8 received from the basis mask sequence generator 820 and generates 32 code symbols being all 0s. The adder 860 adds the code symbols received from the multipliers 840 to 849 and outputs final code symbols "0100000100001000110011011100001". The final code symbols can be achieved by adding the basis Walsh codes W1, W2, W4 and W16 corresponding to the information bits 1s to the basis mask sequence M1 symbol by symbol. In other words, the basis Walsh codes W1, W2, W4 and W16 are summed to W23 and the Walsh code W23 and the basis mask sequence M1 are added to form the TFCI codeword (final code symbols) (=W23+M1) which is outputted from the adder 860.

FIG. 11 is a flowchart illustrating an embodiment of a TFCI encoding procedure in an IMT 2000 system according to the present invention.

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Referring to FIG. 11, 10 input information bits (i.e., TFCI bits) are received and variables sum and j are set to an initial value 0 in step 1100. The variable sum indicates final code symbols, and j indicates the count number of final code symbols output after symbol-basis addition. In step 1110, it is determined whether j is 32 in view of the length 32 symbols of Walsh codes and mask sequences used for encoding the input information bits. Step 1110 is performed in order to check whether the input information bits are all encoded with the Walsh codes and the mask sequences symbol by symbol.

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If j is not 32 in step 1110, which implies that the input information bits are not encoded completely with respect to all symbols of the Walsh codes, the mask sequences, jth symbols W1(j), W2(j), W4(j), W8(j), and W16(j) of the basis Walsh codes W1, W2, W4, W8, and W16 and jth symbols M1(j), M2(j), M4(j), and M8(j) of the basis mask sequences M1, M2, M4, and M8 are received in step 1120. Then, the received symbols are multiplied by the input information bits on a symbol basis and the symbol products are summed in step 1130. The sum becomes the variable sum.

Step 1130 can be expressed as

```
sum = a0 + a1 \cdot W1(j) + a2 \cdot W2(j) + a3 \cdot W4(j) + a4 \cdot W8(j) + a5 \cdot W16(j) + a6 \cdot M1(j) + a7 \cdot M2(j) + a8 \cdot M4(j) + a9 \cdot M8(j)  .... (Eq. 9)
```

As noted from Eq. 9, the input information bits are multiplied by corresponding symbols of the basis Walsh codes and basis mask sequences, symbol products are summed, and the sum becomes an intended code symbol.

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In step 1140, sum indicating the achieved jth code symbol, is output. j is increased by 1 in step 1150 and then the procedure returns to step 1110. Meanwhile, if j is 32 in step 1110, the encoding procedure ends.

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The encoding apparatus of FIG. 8 according to the embodiment of the present invention can support extended TFCIs as well as basic TFCIs. Encoders for supporting an extended TFCI include a (32, 10) encoder, a (32, 9) encoder, and a (32, 7) encoder.

combination of 32 Walsh codes of length 32, 32 bi-orthogonal codes inverted from the Walsh codes, and 15 mask sequences. The 32 Walsh codes can be generated from

combinations of 5 basis Walsh codes. The 32 bi-orthogonal codes can be obtained by adding 1 to the 32 symbols of each Walsh code. This results has the same effect as multiplication of -1 by the 32 Walsh codes viewed as real numbers. The 15 mask sequences can be achieved through combinations of 5 basis mask sequences. Therefore,

For the input of 10 input information bits, the (32, 10) encoder outputs a

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The (32, 9) encoder receives 9 input information bits and outputs a combination of 32 Walsh codes of length 32, 32 bi-orthogonal codes inverted from the Walsh codes, and 4 mask sequences. The 4 mask sequences are obtained by combing two of 4 basis mask sequences.

a total of 1024 codewords can be produced from the (32, 10) encoder.

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The (32, 7) encoder receives 7 input information bits and outputs a combination of 32 Walsh codes of length among the 1024 codewords, 32 bi-orthogonal codes inverted from the Walsh codes, and one of 4 basis mask sequences.

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The above encoders for providing extended TFCIs have a minimum distance 12 and can be implemented by blocking input and output of at least of the 4 basis mask sequences generated from the basis mask sequences 820.

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That is, the (32, 9) encoder can be implemented by blocking input and output of one of the four basis mask sequences generated from the basis mask sequence generator 820 shown in FIG. 8. The (32, 8) encoder can be implemented by blocking input and

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output of two of the basis mask sequences generated from the basis mask sequence generator 820. The (32, 7) encoder can be implemented by blocking input and output of three of the basis mask sequences generated from the basis mask sequence generator 820. As described above, the encoding apparatus according to the embodiment of the present invention can encode flexibly according to the number of input information bits, that is, the number of TFCI bits to be transmitted and maximizes a minimum distance that determined the performance of the encoding apparatus.

Codewords in the above encoding apparatus are sequences obtained by combining 32 Walsh codes of length 32, 32 bi-orthogonal codes resulting from adding 1s to the Walsh codes, and 15 mask sequences of length 15. The structure of the codewords is shown in FIG. 13.

For better understanding of the TFC bits encoding procedure, Tables 1a to 1f list code symbols (TFCI codewords) versus 10 TFCI bits.

(Table 1a)

(14010 14)			
000000000	:	000000001	:
000000000000000000000000000000000000000		111111111111111111111111111111111111111	
000000010	:	000000011	:
0101010101010101010101010101010101		10101010101010101010101010101010	
000000100	:	000000101	:
00110011001100110011001100110011		11001100110011001100110011001100	
000000110	:	0000000111	:
01100110011001100110011001100110	•	10011001100110011001100110011001	
0000001000	:	0000001001	:
00001111000011110000111100001111		11110000111100001111000011110000	
0000001010	:	0000001011	:
01011010010110100101101001011010		10100101101001011010010110100101	
0000001100	:	0000001101	:
00111100001111000011110000111100		11000011110000111100001111000011	
0000001110	:	0000001111	:
01101001011010010110100101101001		10010110100101101001011010010110	
0000010000	:	0000010001	:
00000000111111110000000011111111		111111110000000011111111100000000	
0000010010	:	0000010011	:

01010101101010100101010110101010		1010101001010101010101010101010101	
0000010100	:	0000010101	:
00110011110011000011001111001100		11001100001100111100110000110011	
0000010110	:	0000010111	:
01100110100110010110011010011001		10011001011001101001100101100110	
0000011000	:	0000011001	:
0000111111111000000001111111110000		111100000000111111111000000001111	
0000011010	:	0000011011	:
010110101010010101011010101010101		10100101010110101010010101011010	
0000011100	:	0000011101	:
00111100110000110011110011000011		11000011001111001100001100111100	
0000011110	:	0000011111	:
01101001100101100110100110010110		10010110011010011001011001101001	
0000100000	:	0000100001	:
0000000000000001111111111111111111		111111111111111111000000000000000000	
0000100010	:	0000100011	:
01010101010101011010101010101010		10101010101010100101010101010101	
0000100100	:	0000100101	:
00110011001100111100110011001100		11001100110011000011001100110011	
0000100110	:	0000100111	:
01100110011001101001100110011001		10011001100110010110011001100110	
0000101000	:	0000101001	:
000011110000111111111000011110000		11110000111100000000111100001111	
0000101010	:	0000101011	:
01011010010110101010010110100101		10100101101001010101101001011010	
0000101100	:	0000101101	:
00111100001111001100001111000011		11000011110000110011110000111100	
0000101110	:	0000101111	:
01101001011010011001011010010110		10010110100101100110100101101001	
0000110000	:	0000110001	:
000000001111111111111111100000000		11111111000000000000000011111111	
0000110010	:	0000110011	:
01010101101010101010101001010101		10101010010101010101010110101010	
0000110100	:	0000110101	:
00110011110011001100110000110011		11001100001100110011001111001100	
0000110110	:	0000110111	:

1000110100111001010101010100101101 0111001011000110101010101011010010 0001011100: 0001011101: 00010100101000001100110010110100 111010110101111110011001101001011 0001011110: 0001011111: 101111110000010100110011000011110 010000011111010110011001111100001 0001100001: 0001100000: 00101000011000110000111110001000 11010111100111001111000001110111 0001100011: 0001100010: 1000001011001001101001010010010 01111101001101100101101011011101 0001100100: 0001100101: 111001001010111111100001101000100 00011011010100000011110010111011 0001100111: 0001100110: 01001110000001010110100111101110 101100011111110101001011000010001 0001101001: 0001101000: 11011000100100111111111111101111000 00100111011011000000000010000111 0001101010: 0001101011: 10001101110001101010101000101101 011100100011100101010101111010010 0001101101: 0001101100: 000101000101111110011001110110100 11101011101000001100110001001011 0001101111: 0001101110: 10111110111101011001100100011110 01000001000010100110011011100001 0001110000: 0001110001: 110101110110001111111000010001000 00101000100111000000111101110111 0001110011: 0001110010: 10000010001101101010010111011101 01111101110010010101101000100010 0001110101: 0001110100: 11100100010100001100001110111011 00011011101011110011110001000100 0001110111: 0001110110: 01001110111110100110100100010001 10110001000001011001011011101110 0001111001: 0001111000: 00100111100100110000000001111000 1101100001101100111111111110000111 0001111010: 0001111011: 1000110100111001101010101101010 0111001011000110010101010101101 0001111101: 0001111100: 00010100101000000011001101001011 111010110101111111100110010110100 0001111110: 0001111111:

10111110000010101001100111100001 01000001111101010110011000011110 0010000001: 0010000000: 000000011100110101101101111000111 11111110001100101001001000111000 0010000010: 0010000011: 101010110110011111100011101101101-01010100100110000011100010010010 0010000101: 0010000100: 11001101000000011010000100001011 0011001011111111001011111011110100 0010000110: 0010000111: 10011000010101001111010001011110 01100111101010110000101110100001 0010001001: 0010001000: 00001110110000100110001011001000 11110001001111011001110100110111 0010001011: 0010001010: 10100100011010001100100001100010 01011011100101110011011110011101 0010001100: 0010001101: 110000100000111010101111000000100 0011110111111000101010001111111011 0010001110: 0010001111: 01101000101001000000010010101110 100101110101101111111101101010001 0010010001: 0010010000: 11111110110011011001001011000111 00000001001100100110110100111000 0010010011: 0010010010: 10101011100110001100011110010010 01010100011001110011100001101101 0010010101: 0010010100: 1100110111111111010100001111110100 001100100000000101011111000001011 0010010111: 0010010110: 100110001010101111111010010100001 01100111010101000000101101011110 0010011001: 0010011000: 11110001110000101001110111001000 00001110001111010110001000110111 0010011011: 0010011010: 10100100100101111100100010011101 01011011011010000011011101100010 0010011101: 0010011100: 00111101000011100101000100000100110000101111000110101111011111011 0010011111: 0010011110: 0110100001011011000001000101000110010111101001001111101110101110 0010100001: 0010100000: 1111111100011001001101101111000111 00000001110011011001001000111000 0010100011: 0010100010:

01010100100110001100011101101101 10101011011001110011100010010010 0010100100: 0010100101: 001100101111111101010000100001011 110011010000000101011111011110100 0010100110: 0010100111: 10011000010101000000101110100001 011001111010101111111010001011110 0010101000: 0010101001: 00001110110000101001110100110111 11110001001111010110001011001000 0010101010: 0010101011: 10100100011010000011011110011101 010110111001011111100100001100010 0010101101: 0010101100: 11000010000011100101000111111011 0011110111111000110101111000000100 0010101110: 0010101111: 011010001010010011111101101010001 100101110101101100000100101011110 0010110001: 0010110000: 11111110110011010110110100111000 00000001001100101001001011000111 0010110010: 0010110011: 010101000110011111000111110010010 10101011100110000011100001101101 0010110100: 0010110101: 00110010000000011010000111110100 1100110111111111001011111000001011 0010110110: 0010110111: 10011000101010110000101101011110 011001110101010011111010010100001 0010111000: 0010111001: 11110001110000100110001000110111 00001110001111011001110111001000 0010111011: 0010111010: 1010010010010111100110111101100010 01011011011010001100100010011101 0010111101: 0010111100: 11000010111100010101000100000100 001111010000111010101111011111011 0010111111: 0010111110: 011010000101101111111101110101110 10010111101001000000010001010001 0011000001: 0011000000: 11010110010100010110001001001111 001010011010111010011101101101000 0011000010: 0011000011: 10000011000001000011011100011010 01111100111110111100100011100101 0011000101: 0011000100: 00011010100111011010111010000011 111001010110001001010001011111100 0011000111: 0011000110:

01001111110010001111101111010110	10110000001101110000010000101001
0011001000 :	
00100110101000011001001010111111	·

(Table 1b)

	(1able 10)	
	0011001001:	0011001010:
	110110010101111100110110101000000	011100111111010011000111111101010
	0011001011:	0011001100:
	10001100000010110011100000010101	00010101100100101010000110001100
	0011001101:	0011001110:
	111010100110110101011111001110011	010000001100011111111010011011001
	0011001111:	0011010000:
	10111111001110000000101100100110	00101001010100011001110101001111
	0011010001:	0011010010:
	11010110101011100110001010110000	01111100000001001100100000011010
	0011010011:	0011010100:
	100000111111101100110111111100101	000110100110001010101111001111100
-	0011010101:	0011010110:
	11100101100111010101000110000011	010011110011011111111101100101001
	0011010111:	0011011000:
	10110000110010000000010011010110	00100110010111101001001001000000
	0011011001:	0011011010:
	11011001101000010110110110111111	01110011000010111100011100010101
	0011011011 :	0011011100:
	10001100111101000011100011101010	00010101011011011010000101110011
	0011011101 :	0011011110:
	11101010100100100101111010001100	01000000001110001111010000100110
	0011011111:	0011100000:
	10111111110001110000101111011001	00101001101011100110001001001111
	0011100001 :	0011100010:
	11010110010100011001110110110000	011111001111101100110111100011010
	0011100011:	0011100100:
	10000011000001001100100011100101	000110101001110101010001011111100
	0011100101:	0011100110:
	111001010110001010101111010000011-	01001111110010000000010000101001

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0101111101 :	0101111110:
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0110000011:	0110000100:
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0110000111:	0110001000:
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0110001001 :	0110001010:
11111011110001001000011000011100	01010001011011100010110010110110
0110001011:	0110001100:
10101110100100011101001101001001	00110111000010000100101011010000
0110001101:	0110001110:
11001000111101111011010100101111	01100010010111010001111110000101
0110001111:	0110010000:
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0110010001 :	
11110100001101001000100111101100	

(Table 1c)

0110010010:	0110010011:	ı
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(Table 1d)

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	1001011111:	1001100000:
	10100010001111010100100101001111	0110100010101000010000011011001
	1001100001:	1001100010 :
	1100101110101011111011111100100110	1100001000000010111010110001100
į	1001100011:	1001100100:
	100111101111111101000101001110011	000001110110011100010011111101010
	1001100101 :	1001100110 :
	1111000100110001110110000010101	01010010001100100100011010111111
	1001100111:	1001101000 :
	10101101110011011011100101000000	001110110101101100101111111010110
	1001101001:	1001101010:
~	1000100101001001101000000101001	01101110000011100111101010000011

1001101100: 1001101011: 100100011111000110000101011111100 0001000011010000001110011100101 1001101110: 1001101101: 111101111001011111110001100011010 01011101001111010100100110110000 1001110000: 1001101111: 0110100101010110010000000100110 10100010110000101011011001001111 1001110010: 1001110001: 0110000111111111001110101011110011 1100101101010100110111111111011001 1001110011:100111110000000011000101 1001110100: 010001100 0000111100110000001001100010101 1001110110: 1001110101: 1010010110011010100011001000000 111110000110011111110110011101010 1001110111: 1001111000: 00111011101001000010111100101001 101011010011001010111100110111111 1001111010: 1001111001: 011011101111000101111101001111100 110001000101101111101000011010110 1001111100: 1001111011: 0010001000011101000010110000011 00001000100101110001110000011010 1001111101 : 1001111110: 11110111011010001110001111100101 01011101110000100100100101001111 1010000000: 1001111111: 00011101111110100100001010010110 1010001000111101101101101101010000 1010000010: 1010000001: 0100100010101111100010111111000011 1100010000001011011110101101001. 1010000100: 1010000011: 0110111010100001110100000111100 01011101100100101111000110100101 1010000110: 1010000101: 011110111001110000100100111110000 1010001001101101000111001011010 1010001000: 1010000111: 0010010111101010100110110011001 10000100011000111101101100001111 1010001010: 1010001001: 1000111101000000001100011001100 1101101000010101011001001100110 1010001100: 1010001011: 1011100001011111111110011100110011 001000011100011001111111010101010 1010001101: 1010001110: 1011110001110011000000101010101 011101001001001100101011111111111

1010010000: 1010001111: 0001011011011001101010000000000 0011101000001010100001001101001 1010010001: 1010010010: 1001000010100000001011100111100 111000101111101010111110110010110 1010010100: 1010010011: 0101110001101100111000101011010 101101111010111111110100011000011 1010010101: 1010010110: 1010001110010011000111010100101 01111011011000110010010000001111 1010011000: 1010010111: 1000010010011100110110111111110000 0010010000010100100110101100110 1010011010: 1010011001: 010001110101111110001100000110011 111011011111101011011001010011001 1010011100: 1010011011: 10111000101000001110011111001100 0010000100111001011111110010101011010011110: 1010011101: 1011110110001101000000110101010 01110100011011000010101100000000 1010100000: 1010011111: 10001011100100111101010011111111 0001110111111101010111110101101001 1010100001: 1010100010: 1100010000001010100001010010110 010010001010111111110100000111100 1010100100: 1010100011: 0101110110010011000111001011010 0110111010100000001011111000011 1010100110: 1010100101: 0111101110011100110110110100001111 1010001001101100111000110100101 1010101000: 1010100111: 0000100011000110010010011110000 0010010111101011011001001100110 1010101010: 1010101001: 01000111101000001110011100110011 1101101000010100100110110011001 1010101100: 1010101011: 01110000101111110001100011001100 0100001110001101000000101010101 1010101110: 1010101101: 110111100011100101111111010101010 11101001001001111010100000000000 1010110000: 1010101111: 100010110110110000101011111111111 00011101000001011011110110010110 1010110001: 1010110010: 1100010111110100100001001101001 1001000010100001110100011000011

1010110011: 1010110100: 0101110001101101000111010100101 1011011110101111100010111100111100 1010110101: 1010110110: 011110110110001111011011111110000 1010001110010010111000101011010 1010110111: 1010111000: 0000100100111000010010000001111 0010010000010101011001010011001 1010111001: 1010111010: 111011011111101010100110101100110 0100011101011111111100111111001100 1010111100: 1010111011: 0111000101000000001100000110011 0100001001110011000000110101010 1010111110: 1010111101: 110111101100011001111111001010101 011101000110110011010101011111111 1011000000: 1010111111: 00010111001001100101011100000000 00110101100110011011001011100001 1011000001: 1011000010: 1001010011001100100110100011110 01100000110011001110011110110100 1011000100: 1011000011: 00001101010101010000001110100100011111001100110001100001001011 1011000101: 1011000110: 010100111111111111101010010000111 111110010101010101111111000101101 1011000111: 1011001000: 0101100000000000010101101111000001110101001011010111110111101110 1011001010: 1011001001: 01101111111000011111101000101111011 1000101011010010100001000010001 1011001011: 1011001100: 00001001101001011000111011011101 0010000001111000001011101000100 1011001110: 1011001101: 111011001011010011100010010010 010111001111000011011011100010001011010000: 1011001111: 10100011000011110010010001110111 00110101011001101011001000011110 1011010010: 1011010001: 01100000001100111110011101001011 10010101001100101001101111100001 1011010100: 1011010011: 0000110010101011000000100101101 100111111110011000001100010110100 1011010101: 1011010110: 1010011000000001101010001111000 111110011010101001111111011010010

1011011000 : 1011010111: 1010110011111111100101011110000111 00111010011010011011110100010001 1011011010: 1011011001: 01101111001111001110100001000100 1000101100101100100001011101110 1011011100: 1011011011: 10010000110000110001011110111011 0001001010110101000111000100010 1011011110: 1011011101: 111101101010010101111000111011101 01011100000011111101101101111111 1011100000: 1011011111: 00110101100110010100110100011110101000111111100000010010010001000 1011100001: 1011100010: 01100000110011000001100001001011 11001010011001101011001011100001 1011100100: 1011100011: 000001101010101001111111000101101 100111110011001111110011110110100 1011100110: 1011100101: 01010011111111111001010110111111000 11111001010101011000000111010010 1011101000: 1011100111: 00111010100101100100001000010001 1010110000000000110101010010000111 1011101010: 1011101001: 11000101011010011011110111101110 0110111111100001100010111101000100 1011101100: 1011101011: 0000100110100101011100010010010 10010000001111001110100010111011 1011101110: 1011101101: 01011100111100000010010001110111 11110110010110101000111011011101 1011110000: 1011101111: 001101010110011001001101111100001 101000110000111111011011110001000 1011110010: 1011110001: 01100000001100110001100010110100 11001010100110011011001000011110 1011110100: 1011110011: 10011111110011001110011101001011 000001100101010101111111011010010 1011110110: 1011110101: 11111001101010101000000100101101 010100110000000000101011110000111 10111111000: 1011110111: 00111010011010010100001011101110 101011001111111111101010001111000 1011111010: 1011111001: 01101111001111000001011110111011 110001011001011010111110100010001 -

1011111011 :	1011111100 :
10010000110000111110100001000100	00001001010110100111000111011101
1011111101 :	1011111110:
11110110101001011000111000100010	01011100000011110010010010001000
1011111111:	1100000000:
10100011111100001101101101110111	00010110110011100011010001111010
1100000001:	1100000010:
11101001001100011100101110000101	010000111001101101100001001011111
1100000011:	1100000100:
10111100011001001001111011010000	00100101111111010000011101001001
1100000101:	1100000110:
11011010000000101111100010110110	01110000101010000101001000011100
1100000111:	1100001000:
100011110101011110101101111100011	00011001110000010011101101110101
1100001001:	1100001010:
11100110001111101100010010001010	01001100100101000110111000100000
1100001011:	1100001100:
10110011011010111001000111011111	00101010111100100000100001000110 -
1100001101:	1100001110:
11010101000011011111011110111001	01111111101001110101110100010011
1100001111 :	1100010000 :
10000000010110001010001011101100	00010110001100010011010010000101
1100010001 :	1100010010:
11101001110011101100101101111010	01000011011001000110000111010000
1100010011:	1100010100:
10111100100110111001111000101111	0100101000000100000011110110110
1100010101 :	1100010110 :
1101101011111110111111100001001001	01110000010101110101001011100011
1100010111 :	1100011000 :
10001111101010001010110100011100	00011001001111100011101110001010
1100011001:	1100011010:
11100110110000011100010001110101	01001100011010110110111011011111
1100011011 :10110011100101001001000	1100011100:
100100000	0101010000011010000100010111001
1100011101:	1100011110:
1101010111111001011111011101000110	01111111010110000101110111101100
•	

 1100011111:
 1100100000:

 1000000010100111101000100010011
 00010110110011101100101110000101

 1100100001:
 1100100010:

 11101001001100010011010001111010
 010000111001101110011110111001000

 1101111000110011001000110000100101111
 0100001110011011100111100100000

(Table 1e)

1100100100:		1100100101:
00100101111	1110111111100010110110	11011010000000100000011101001001
1100100110:		1100100111:
01110000101	010001010110111100011	10001111010101110101001000011100
1100101000:		1100101001:
00011001110	0000011100010010001010	11100110001111100011101101110101
1100101010:	·	1100101011:
01001100100	0101001001000111011111	10110011011010110110111000100000
1100101100:		1100101101:
00101010111	100101111011110111001	11010101000011010000100001000110
1100101110:		1100101111:
01111111101	001111010001011101100	10000000010110000101110100010011
1100110000:	•	1100110001:
00010110001	1000111001101101111010	11101001110011100011010010000101
1100110010:		1100110011:
01000011011	.001001001111000101111	10111100100110110110000111010000
1100110100:		1100110101:
00100101000	0000101111100001001001	110110101111111010000011110110110
1100110110:		1100110111:
01110000001	0101111010110100011100	100011111010100001010010111100011
1100111000:		1100111001:
00011001001	111101100010001110101	11100110110000010011101110001010
1100111010:		1100111011:
01001100011	1010111001000100100000	10110011100101000110111011011111
1100111100:		1100111101:
00101010000	0011011111011101000110	110101011111100100000100010111001
1100111110:		1100111111:
		· · · · · · · · · · · · · · · · · · ·

10000000101001110101110111101100 01111111010110001010001000010011 1101000001: 1101000000: 001111101010110111100010000001101 110000010101001000111011111110010 1101000010: 1101000011: 100101000000011101101111010100111 0110101111111100010010010101011000 1101000101: 1101000100: 11110010011000010000100011000001 00001101100111101111011100111110 1101000111: 1101000110: 10100111001101000101110110010100 01011000110010111010001001101011 1101001000: 1101001001: 11001110010111010011010011111101 00110001101000101100101100000010 1101001011: 1101001010: 011001001111011110011110010101111 10011011000010000110000110101000 1101001101: 1101001100: 1111110101101111000000111111001110 0000001010010001111111000001100011101001110: 1101001111: 10101000001110110101001010011011 0101011111000100101011010110100100 1101010001: 1101010000: 0011111001010010111000100111110010 11000001101011010011101100001101 1101010010: 1101010011: 100101001111110000110111001011000 01101011000001111001000110100111 1101010101: 1101010100: 00001101011000011111011111000001 111100101001111100000100000111110 1101010111: 1101010110: 101001111100101101011110101101011 01011000001101001010001010010100 1101011000: 1101011001: 11001110101000100011010000000010 0011000101011101110010111111111101 1101011011: 1101011010: 1001101111111011101100001010101111 01100100000010001001111010101000 1101011101: 1101011100: 00000010011011101111100011001110 11111101100100010000011100110001 1101011111: 1101011110: 10101000110001000101001001100100 01010111001110111010110110110011011 1101100000: 1101100001: 001111101010110100111011111110010 11000001010100101100010000001101 1101100011: 1101100010:

011010111111110000110111010100111	10010100000001111001000101011000
1101100100:	1101100101:
00001101100111100000100011000001	11110010011000011111011100111110
1101100110:	1101100111:
01011000110010110101110110010100	10100111001101001010001001101011
1101101000:	1101101001:
00110001101000100011010011111101	11001110010111011100101100000010
1101101010:	1101101011:
01100100111101110110000110101000	10011011000010001001111001010111
1101101100:	1101101101:
00000010100100010000011111001110	11111101011011110111111000000110001
1101101110:	1101101111:
01010111110001000101001010011011	10101000001110111010110101100100
1101110000:	1101110001:
00111110010100100011101100001101	11000001101011011100010011110010
1101110010:	1101110011:
01101011000001110110111001011000	100101001111110001001000110100111
1101110100:	1101110101:
00001101011000010000100000111110	11110010100111101111011111000001
1101110110:	1101110111:
01011000001101000101110101101011	10100111110010111010001010010100
1101111000:	1101111001:
00110001010111010011010000000010	110011101010001011001011111111101
1101111010:	1101111011:
01100100000010000110000101010111	10011011111101111001111010101000
1101111100:	1101111101:
00000010011011100000011100110001	111111011001000111111100011001110
1101111110:	1101111111:
01010111001110110101001001100100	101010001100010010110110110011011
1110000000:	1110000001:
00010111000000110101100110111101	11101000111111001010011001000010
1110000010:	1110000011:
01000010010101100000110011101000	101111011010100111111001100010111
1110000100:	1110000101:
00100100001100000110101010001110	110110111100111110010101011110001
1110000110:	1110000111:

011100010110010100111111111011011 10001110100110101100000000100100 1110001001: 1110001000: 00011000000011000101011010110010 -111001111111100111010100101001101 1110001011: 1110001010: 01001101010110010000001111100111 101100101010011011111110000011000 1110001100: 1110001101: 1101010011000000100110100111110 00101011001111110110010110000001 1110001111: 1110001110: 01111110011010100011000011010100 100000011001010111100111100101011 1110010000: 1110010001: 00010111111111000101100101000010 1110100000000111010011010111101 1110010010: 1110010011: 01000010101010010000110000010111 1011110101010110111110011111101000 1110010100: 1110010101: 00100100110011110110101001110001 11011011001100001001010110001110 1110010111: 1110010110: 01110001100110100011111110010010010001110011001011100000011011011 1110011001: 1110011000: 11100111000011001010100110110010 0001100011110011010101101001001101 1110011011: 1110011010: 01001101101001100000001100011000 101100100101100111111110011100111 1110011101: 1110011100: 110101000011111111001101010000001 00101011110000000110010101111110 1110011110: 1110011111: 01111110100101010011000000101011 100000010110101011001111111010100 1110100001: 1110100000: 00010111000000111010011001000010 111010001111111000101100110111101 1110100011: 11101000010: 01000010010101101111001100010111 10111101101010010000110011101000 1110100100: 1110100101: 001001000011000010010101011110001 11011011110011110110101010001110 1110100111: 1110100110: 100011101001101000111111111011011 01110001011001011100000000100100 1110101001: 1110101000: 111001111111100110101011010110101 0001100000001100101010010100101 11101010111: 1110101010:

10110010101001100000001111100111 0100110101011001111111110000011000 1110101100: 1110101101: 11010100110000000110010110000001 0010101100111111110011010011111110 1110101110: 1110101111: 011111100110101011001111100101011 10000001100101010011000011010100 1110110001: 1110110000: 1110100000000110101100101000010 0001011111111110010100110101111101 1110110011: 1110110010: 10111101010101100000110000010111 01000010101010011111001111101000 1110110100: 1110110101: 001001001100111110010101110001110 11011011001100000110101001110001 1110110111: 1110110110: 100011100110010100111111100100100 01110001100110101100000011011011 1110111001: 1110111000: 11100111000011000101011001001101 000110001111001110101001101100101110111010: 1110111011: 10110010010110010000001100011000 010011011010011011111110011100111 11101111100: 1110111101: 0010101111000000100110101010000001 110101000011111101100101011111110 1110111110: 1110111111: 10000001011010100011000000101011 0111111010010101111001111111010100 1111000001:110000001001111101010110 1111000000: 00110101 00111111011000001010100111001010 1111000011: 1111000010: 1001010111001010000000110110000 011010100011010111111110010011111 1111000100: 1111000101: 11110011101011000110010100000110 000011000101001110011010111111001 1111000111: 1111000110: 101001101111110010011000001010011 010110010000011011001111110101100 1111001001: 1111001000: 11001111100100000101100100111010 00110000011011111010011011000101 1111001010: 1111001011: 10011010110001010000110001101111 01100101001110101111001110010000 1111001101: 1111001100: 11111100101000110110101000001001 000000110101110010010101111110110 1111001111: 1111001110:

01010110000010011100000010100011 1010100111110110000111111101011100 1111010000: 1111010001: 001111111001111111010100100110101 11000000011000000101011011001010 1111010011: 1111010010: 10010101001101010000001110011111 011010101100101011111110001100000 1111010101: 1111010100: 111100110101001101100101111111001 00001100101011001001101000000110 1111010111: 1111010110: 01011001111110011100111101010011 10100110000001100011000010101100 1111011001: 1111011000: 11001111011011110101100111000101 00110000100100001010011000111010 1111011011: 1111011010: 011001011100010111111001101101111 10011010001110100000110010010000 1111011100: 1111011101: 111111000101110001101010111110110 00000011101000111001010100001001 1111011111: 1111011110: 1010100100001001001111111110100011 010101101111011011000000011100 1111100001: 1111100000: 110000001001111110101001111001010 00111111011000000101011000110101 1111100010: 1111100011: 100101011100101011111110010011111 01101010001101010000001101100000 1111100101: 1111100100: 111100111010110010011010111111001 00001100010100110110010100000110 1111100111: 1111100110: 10100110111110011100111110101100 01011001000001100011000001010011 1111101001: 1111101000: 11001111100100001010011011000101 00110000011011110101100100111010 1111101011: 1111101010: 100110101100010111111001110010000 01100101001110100000110001101111 1111101100: 00000011010111000110101000001001

(Table 1f)

1111101101

111111001010001110010101111110110

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The decoding apparatus according to the embodiment of the present invention will be described referring to FIG. 9. An input signal r(t) is applied to 15 multipliers 902 to 906 and a correlation calculator 920. The input signal r(t) was encoded with a predetermined Walsh code and a predetermined mask sequence in a transmitter. A mask sequence generator 910 generates all possible 15 mask sequences M1 to M15. The multipliers 902 to 906 multiply the mask sequences received from the mask sequence generator 910 by the input signal r(t). The multiplier 902 multiplies the input signal r(t) by the mask sequence M1 received from the mask sequence generator 910. multiplier 904 multiplies the input signal r(t) by the mask sequence M2 received from the mask sequence generator 910. The multiplier 906 multiplies the input signal r(t) by the mask sequence M15 received from the mask sequence generator 910. If the transmitter encoded TFCI bits with the predetermined mask sequence, one of the outputs of the multipliers 902 to 906 is free of the mask sequence, which means the mask sequence has no effect on the correlations calculated by one of the correlation calculators. For example, if the transmitter used the mask sequence M2 for encoding the TFCI bits, the output of the multiplier 904 that multiplies the mask sequence M2 by the input signal r(t) is free of the mask sequence. The mask sequence-free signal is TFCI bits encoded with the predetermined Walsh code. Correlation calculators 920 to 926 calculate the correlations of the input signal r(t) and the outputs of the multipliers 902 to 906 to 64 bi-orthogonal codes. The 64 bi-orthogonal codes have been defined before. The correlation calculator 920 calculates the correlation values of the input signal r(t) to the 64 bi-orthogonal codes of length 32, selects the maximum correlation value from the 64 correlations, and outputs the selected correlation value, a bi-orthogonal code index corresponding to the selected correlation value, and its unique index "0000" to a correlation comparator 940.

The correlation calculator 922 calculates the correlation values of the output of the multiplier 902 to the 64 bi-orthogonal codes, selects the maximum value of the 64 correlations, and outputs the selected correlation value, a bi-orthogonal code index corresponding to the selected correlation, and its unique index "0001" to the correlation comparator 940. The correlation calculator 924 calculates the correlation values of the output of the multiplier 904 to the 64 bi-orthogonal codes, selects the maximum of the 64 correlation values, and outputs the selected correlation value, a bi-orthogonal code index corresponding to the selected correlation value, and its unique index "0010" to the correlation comparator 940. Other correlation calculators(not shown) calculate the correlation values of the outputs of the correspondent multipliers to the 64 bi-orthogonal codes and operate similar to the above described correlation calculators, respectively.

Finally, the correlation calculator 926 calculates the correlation values of the output of the multiplier 906 to the 64 bi-orthogonal codes, selects the maximum value of the 64 correlations, and outputs the selected correlation value, a bi-orthogonal code index corresponding to the selected correlation value, and its unique index "1111" to the correlation comparator 940.

The unique indexes of the correlation calculators 920 to 926 are the same as the indexes of the mask sequences multiplied by the input signal r(t) in the multipliers 902 to 906. Table 2 lists the 15 mask indexes multiplied in the multipliers and a mask index assigned to the case that no mask sequence is used, by way of example.

(Table 2)

mask sequence	mask sequence index	mask sequence	mask sequence index
not used	0000	M8	1000
M1	0001	M9	1001
M2	0010	M10	1010
M3	0011	M11	1011
M4	0101	M12	1100
M5	0101	M13	1101
M6	0110	M14	1110
M7 .	0111	M15	1111

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As shown in Table 2, the correlation calculator 922, which receives the signal which is the product of the input signal r(t) and the mask sequence M1, outputs "0001" as its index. The correlation calculator 926, which receives the signal which is the product of the input signal r(t) and the mask sequence M15, outputs "1111" as its index. The correlation calculator 920, which receives only the input signal r(t), outputs "0000" as its index.

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Meanwhile, the bi-orthogonal code indexes are expressed in a binary code. For example, if the correlation to $\overline{W4}$ which is the complement of W4is the largest correlation value, a corresponding bi-orthogonal code index (a0 to a9) is "001001".

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The correlation comparator 940 compares the 16 maximum correlation values received from the correlation calculators 920 to 926, selects the highest correlation value

from the 16 received maximum correlation values, and outputs TFCI bits based on the bi-orthogonal code index and the mask sequence index(the unique index) received from the correlation calculator that corresponds to the highest correlation value. The TFCI bits can be determined by combining the bi-orthogonal code index and the mask sequence index. For example, if the mask sequence index is that of M4(0100) and the bi-orthogonal code index is that of $\overline{W4}$ (001001), the TFCI bits(a9 to a0) are "the M4 index(0100) + the $\overline{W4}$ index(001001)". That is, the TFCI bits(a9 to a0) are "0100001001"

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Assuming that the transmitter transmitted code symbols corresponding to TFCI bits (a0 to a9) "1011000010", it can be said that the transmitter encoded the TFCI bits with $\overline{W6}$ and M4 according to the afore-described encoding procedure. The receiver can determine that the input signal r(t) is encoded with the mask sequence M4 by multiplying the input signal r(t) by all the mask sequences and that the input signal r(t) is encoded with $\overline{W6}$ by calculating the correlations of the input signal r(t) to all the biorthogonal codes. Based on the above example, the fifth correlation calculator(not shown) will output the largest correlation value, the index of $\overline{W6}$ (101100) and its unique index(0010). Then, the receiver outputs the decoded TFCI bits(a0 to a9) "1011000010" by adding the index of $\overline{W6}$ "101100" and the M4 index "0010".

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In the embodiment of the decoding apparatus, the input signal r(t) is processed in parallel according to the number of mask sequences. It can be further contemplated that the input signal r(t) is sequentially multiplied by the mask sequences and the correlations of the products are sequentially calculated in another embodiment of the decoding apparatus.

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FIG. 17 illustrates another embodiment of the decoding apparatus.

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Referring to FIG. 17, a memory 1720 stores an input 32-symbol signal r(t). A mask sequence generator 1710 generates 16 mask sequences that were used in the transmitter and outputs them sequentially. A multiplier 1730 multiplies one of the 16 mask sequences received from the mask sequence generator 1710 by the input signal r(t) received from the memory 1720. A correlation calculator 1740 calculates the output of the multiplier 1730 to 64 biorthogonal codes bi-orthogonal of length 32 and outputs the maximum correlation value and the index of a biorthogonal code corresponding to the largest correlation value to a correlation comparator 1750. The correlation comparator

1750 stores the maximum correlation value and the biorthogonal code index received from the correlation calculator 1740, and the index of the mask sequence received from the mask sequence generator 1710.

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Upon completion of above processing with the mask sequence, the memory 1720 outputs the stored input signal r(t) to the multiplier 1730. The multiplier 1730 multiplies the input signal r(t) by one of the other mask sequences. The correlation calculator 1740 calculates correlation of the the output of the multiplier 1730 to the 64 biorthogonal codes of length 32 and outputs the maximum correlation value and the index of a biorthogonal code corresponding to the maximum correlation value. The correlation comparator 1750 stores the maximum correlation value, the biorthogonal code index corresponding to the maximum correlation value, and the mask sequence index received from the mask sequence generator 1710.

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The above procedure is performed on all of the 16 mask sequences generated from the mask sequence generator 1710. Then, 16 maximum correlation values the indexes of biorthogonal codes corresponding to the maximum correlation value are stored in the correlation comparator 1750. The correlation comparator 1750 compares the stored 16 correlation values and selects the one with the highest correlation and outputs TFCI bits by combining the indexes of the biorthogonal code and mask sequence index corresponding to the selected maximum correlation value. When the decoding of the TFCI bits is completed, the input signal r(t) is deleted from the memory 1720 and the next input signal r(t+1) is stored.

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While the correlation comparator 1750 compares the 16 maximum correlation values at one time in the decoding apparatus of FIG. 17, real-time correlation value comparison can be contemplated. That is, the first input maximum correlation value is compared with the next input maximum correlation value and the larger of the two correlation values and a mask sequence index and a biorthogonal code index corresponding to the correlation are stored. Then, the thirdly input maximum correlation is compared with the stored correlation and the larger of the two correlations and a mask sequence index and a biorthogonal code index corresponding to the selected correlation are stored. This comparision/operation occurs 15 times which is the number of mask sequences generated from the mask sequence generator 1710. Upon completion of all the operations, the correlation comparator 1710output the finally stored biorthogonal index(a0 to a6) and mask sequence index(a7 to a9) and outputs the added bits as TFCI bits.

FIG. 10 is a flowchart illustrating the operation of the correlation comparator 940 shown in FIG. 9. The correlation comparator 940 stores the sixteen maximum correlation values, selects a highest correlation value out of the 16 maximum correlation values and output TFCI bits based on the indexes of a bi-orthogonal code and a mask sequence corresponding to the selected highest correlation value. The sixteen correlation values are compared, and TFCI bits are outputted based on the indexes of a bi-orthogonal code and a mask sequence corresponding to the highest correlation value.

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Referring to FIG. 10, a maximum correlation index i is set to 1 and the indices of a maximum correlation value, a biorthogonal code, and a mask sequence to be checked are set to 0s in step 1000. In step 1010, the correlation comparator 940 receives a 1st maximum correlation value, a 1st bi-orthogonal code index, and a 1st mask sequence index from the correlation calculator 920. The correlation comparator 940 compares the 1st maximum correlation with an the previous maximum correlation value in step 1020. If the 1st maximum correlation is greater than the previous maximum correlation, the procedure goes to step 1030. If the 1st maximum correlation is equal to or smaller than the previous maximum correlation, the procedure goes to step 1040. In step 1030, the correlation comparator 940 designates the 1st maximum correlation as a final maximum correlation and stores the 1st bi-orthogonal code and mask sequence indexes as final biorthogonal code and mask sequence indexes. In step 1040, the correlation comparator 940 compares the index i with the number 16 of the correlation calculators to determine whether all 16 maximum correlations are completely compared. If i is not 16, the index i is increased by 1 in step 1060 and the procedure returns to step 1010. Then, the above procedure is repeated.

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In step 1050, the correlation comparator 940 outputs the indexes of the biorthogonal code and the mask sequence that correspond to the final maximum correlation as decoded bits. The bi-orthogonal code index and the mask sequence index corresponding to the decoded bits are those corresponding to the final maximum correlation among the 16 maximum correlation values received from the 16 correlation calculators.

3. Second Embodiment of Encoding/Decoding Apparatus and Method

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The (32, 10) TFCI encoder that outputs a 32-symbol TFCI codeword in view of 16 slots has been described in the first embodiment of the present invention. Recently,

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the IMT-2000 standard specification dictates having 15 slots in one frame. Therefore, the second embodiment of the present invention is directed to a (30, 10) TFCI encoder that outputs a 30-symbol TFCI codeword in view of 15 slots. Therefore, the second embodiment of the present invention suggests an encoding apparatus and method for outputting 30 code symbols by puncturing two symbols of 32 coded symbols (codeword) as generated from the (32, 10) TFCI encoder.

The encoding apparatuses according to the first and second embodiments of the present invention are the same in configuration except that sequences output from a one-bit generator, a basis Walsh code generator, and a basis mask sequence generator. The encoder apparatus outputs coded symbols of length 30 with symbol #0(1st symbol) and symbol #16(17th symbol) are punctured in the encoding apparatus of the second embodiment.

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Referring to FIG. 8, 10 input information bits a0 to a9 are applied to the input of the 840 to 849. The one-bit generator 800 outputs symbols 1s(length 32) to the multiplier 840. The multiplier 840 multiplies the input information bit a0 by each 32 symbol received from the one-bit generator 800. The basis Walsh code generator 810 simultaneously generates basis Walsh codes W1, W2, W4, W8, and W16 of length 32. The multiplier 841 multiplies the input information bit al by the basisWalsh code W1 The multiplier 842 multiplies the input "01010101010101010101010101010101". code the basis Walsh W2. a2 information bit by The multiplier 843 multiplies the input. "00110011001100110011001100110011". basis Walsh code bit a3 the information bу The multiplier 844 multiplies the input "00001111000011110000111100001111". basis Walshcode W8 the information bit a4 bу The multiplier 845 multiplies the input "0000000011111111100000000111111111". Walsh code W16 bit basis information a5 by the

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The basis mask sequence generator 820 simultaneously generates basis mask sequences M1, M2, M4, and M8 of length 32. The multiplier 846 multiplies the input information bit a6 the basis mask sequence M1 by The multiplier 847 multiplies the input "001010000110001111111000001110111". M2 basis sequence bit the mask . information a7 by The multiplier 848 multiplies the input "000000011100110101101101111000111". basis mask sequence M4 bit a8 by the information ·

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"00001010111110010001101100101011". The multiplier 849 multiplies the input information bit a9 by the basis mask sequence M8 "00011100001101110010111101010001". The multipliers 840 to 849 function like switches that control the output of or the generation of the bits from the one-bit generator, each of the basis walsh codes and each of the basis mask sequences.

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The adder 860 sums the outputs of the multipliers 840 to 849 symbol by symbol and outputs 32 coded symbols (i.e., a TFCI codeword). Out of the 32 coded symbols, two symbols will be punctured at predetermined positions (i.e. the symbol #0(the first symbol) and symbol #16(the 17th symbol) of the adder 860 output are punctured). The remaining 30 symbols will become the 30 TFCI symbols. It will be easy to modify the second embodiment of present invention. For example, the one-bit generator 800, basis walsh generator 810, basis mask sequence generator 820 can generate 30 symbols which excludes the #0 and #16 symbols. The adder 860 then adds the output of the one-bit generator 800, basis walsh generator 810 and basis mask sequence generator 820 bit by bit and output 30 encoded symbols as TFCI symbols.

FIG. 12 is a encoding method for the second embodiment of present invention. The flowchart illustrating the steps of the encoding apparatus according to the second embodiment of the present invention when the number of slots is 15.

Referring to FIG. 12, 10 input information bits a0 to a9 are received and variables sum and j are set to an initial value 0 in step 1200. In step 1210, it is determined whether j is 30. If j is not 30 in step 1210, the jth symbols W1(j), W2(j), W4(j), W8(j), and W16(j) of the basis Walsh codes W1, W2, W4, W8, and W16 (each having two punctured bits) and the jth symbols M1(j), M2(j), M4(j), and M8(j) of the basis mask sequences M1, M2, M4, and M8 (each having two punctured bits) are received in step 1220. Then, the received symbols are multiplied by the input information bits on a symbol basis and the multiplied symbols are summed in step 1230. In step 1240, sum indicating the achieved jth code symbol is output. j is increased by 1 in step 1250 and then the procedure returns to step 1210. Meanwhile, if j is 30 in step 1210, the encoding procedure ends.

The (30, 10) encoder outputs 1024 codewords equivalent to the codewords of the (32, 10) encoder with symbols #0 and #16 punctured. Therefore, the total number of information can be expressed is 1024.

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The output of a (30, 9) encoder is combinations of 32 Walsh codes of length 30 obtained by puncturing symbols #0 and #16 of each of 32 Walsh codes of length 32, 32 bi-orthogonal codes obtained by adding 1 to each symbol of the punctured Walsh codes (by multiplying -1 to each symbol in the case of a real number), and 8 mask sequences obtained by combining any three of the four punctured basis mask sequences.

The output of a (30, 8) encoder is combinations of 32 Walsh codes of length 30 obtained by puncturing #0 and #16 symbols from each of 32 Walsh codes having a length 32 symbols, 32 bi-orthogonal codes obtained by adding 1 to each symbol of the punctured Walsh codes (by multiplying -1 to each symbol in the case of a real number), and 4 mask sequences obtained by combining any two of the four punctured basis mask sequences.

The output of a (30, 7) encoder is combinations of 32 Walsh codes of length 30 obtained by puncturing #0 and #16 symbols from each of 32 Walsh codes having a length 32 symbols, 32 bi-orthogonal codes obtained by adding 1 to each symbol of the punctured Walsh codes (by multiplying -1 to each symbol in the case of a real number), and one of the four punctured basis mask sequences.

All the above encoders for providing an extended TFCI have a minimum distance of 10. The (30, 9), (30, 8), and (30, 7) encoders can be implemented by blocking input and output of at least one of the four basis mask sequences generated from the basis mask sequence generator 820 shown in FIG. 8.

The above encoders flexibly encode TFCI bits according to the number of the TFCI bits and has a maximized minimum distance that determines encoding performance.

A decoding apparatus according to the second embodiment of the present invention is the same in configuration and operation as the decoding apparatus of the first embodiment except for different signal lengths of the encoded symbols. That is, after (32,10) encoding, two symbols out of the 32 encoded symbols are punctured, or basis walsh codes with two punctured symbols and basis mask sequences with two punctured symbols are used for generating the 30 encoded symbols. Therefore, except for the received signal r(t) which includes a signal of 30 encoded symbols and insertion of dummy signals at the punctured positions, all decoding operations are equal to the description of the first embodiment of present invention.

As FIG. 17, this second embodiment of decoding also can be implemented by a single multiplier for multiplying the masks with r(t) and a single correlation calculator for calculating correlation values of bi-orthogonal codes.

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4. Third Embodiment of Encoding/Decoding Apparatus and Method

The third embodiment of the present invention provides an encoding apparatus for blocking the output of a one-bit generator in the (30, 7), (30, 8), (30, 9) or (30, 10) (hereinafter we express (30, 7-10))encoder of the second embodiment and generating another mask sequence instead in order to set a minimum distance to 11. The encoders refer to an encoder that outputs a 30-symbol TFCI codeword for the input of 7, 8, 9 or 10 TFCI bits.

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FIG. 14 is a block diagram of a third embodiment of the encoding apparatus for encoding a TFCI in the IMT 2000 system. In the drawing, a (30, 7-10) encoder is configured to have a minimum distance of 11.

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The encoding apparatus of the third embodiment is similar in structure to that of the second embodiment except that a mask sequence generator 1480 for generating a basis mask sequence M16 and a switch 1470 for switching the mask sequence generator 1480 and a one-bit generator 1400 to a multiplier 1440 are further provided to the encoding apparatus according to the third embodiment of the present invention.

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The two bit punctured basis mask sequences M1, M2, M4, M8, and M16 as used in FIG. 14 are

M1 = 00000101111110000101101001111110

M2 = 000110001100110001111010110111

M4 = 010111100111101010000001100111

M8 = 011011001000001111011100001111

M16 = 1001000111110011111100010101011

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Referring to FIG. 14, when a (30, 6) encoder is used, the switch 1470 switches the one-bit generator 1400 to the multiplier 1440 and blocks all the basis mask sequences generated from a basis mask sequence generator 1480. The multiplier 1440 multiplies the symbols from the one-bit generator 1400 with the input information bit a0, symbol by symbol.

If a (30, 7-10) encoder is used, the switch 1470 switches the mask sequence generator 1480 to the multiplier 1440 and selectively uses four basis mask sequences generated from a basis mask sequence generator 1420. In this case, 31 mask sequences M1 to M31 can be generated by combining 5 basis mask sequences.

The structure and operation of outputting code symbols for the input information bits a0 to a9 using multipliers 1440 to 1449 are the same as the first and second embodiments. Therefore, their description will be omitted.

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As stated above, the switch 1470 switches the mask sequence generator 1480 to the multiplier 1440 to use the (30, 7-10) encoder, whereas the switch 1470 switches the one-bit generator 1400 to the multiplier 1440 to use the (30, 6) encoder.

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For the input of 6 information bits, the (30, 6) encoder outputs a 30-symbol codeword by combining 32 Walsh codes of length 30 with 32 bi-orthogonal codes obtained by inverting the Walsh codes by the use of the one-bit generator 1400.

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For the input of 10 information bits, the (30, 10) encoder outputs a 30-symbol codeword by combining 32 Walsh codes of length 30 and 32 mask sequences generated using five basis mask sequences. Here, the five basis mask sequences are M1, M2, M4, M8, and M16, as stated above and the basis mask sequence M16 is output from the mask sequence generator 1480 that is added for the encoding apparatus according to the third embodiment of the present invention. Hence, 1024 codewords can be achieved from the (30, 10) encoder. The (30, 9) encoder outputs a 30-symbol codeword by combining 32 Walsh codes and 16 mask sequences, for the input of 9 information bits. The 16 mask sequences are achieved by combining four of five basis mask sequences. The (30, 8) encoder outputs a 30-symbol codeword by combining 32 Walsh codes and 8 mask sequences, for the input of 8 information bits. The 8 mask sequences are obtained by combining three of five basis mask sequences. For the input of 7 information bits, the (30, 7) encoder outputs a 30-symbol codeword by combining 32 Walsh codes of length 30 and four mask sequences. The four mask sequences are obtained by combining two of five basis mask sequences.

All the above (30, 7-10) encoders have a minimum distance of 11 to provide extended TFCIs. The (32, 7-10) encoders can be implemented by controlling use of at least one of the five basis mask sequences generated from the basis mask sequence

generator 1420 and the mask sequence generator 1480 shown in FIG. 14.

FIG. 16 is a flowchart illustrating a third embodiment of the TFCI encoding procedure in the IMT 2000 system according to the present invention.

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Referring to FIG. 16, 10 information bits (TFCI bits) a0 to a9 are received and variables sum and j are set to initial values 0s in step 1600. The variable sum indicates a final code symbol output after symbol-basis addition and the variable j indicates the count number of final code symbols output after the symbol-basis addition. It is determined whether j is 30 in step 1610 in view of the length 30 of punctured Walsh codes and mask sequences used for encoding. The purpose of performing step 1610 is to judge whether the input information bits are encoded with respect to the 30 symbols of each Walsh code and the 30 symbols of each mask sequence.

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If j is not 30 in step 1610, which implies that encoding is not completed with respect to all the symbols of the Walsh codes and mask sequences, the jth symbols W1(j), W2(j), W4(j), W8(j), and W16(j) of the basis Walsh codes W1, W2, W4, W8, and W16 and the jth symbols M1(j), M2(j), M4(j), M8(j), and M16(j) of the basis mask sequences M1, M2, M4, M8, and M16 are received in step 1620. In step 1630, the input information bits are multiplied by the received symbols symbol by symbol and the symbol products are summed.

Step 1630 can be expressed as

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 $sum = a0 \cdot M16(j) + a1 \cdot W1(j) + a2 \cdot W2(j) + a3 \cdot W4(j) + a4 \cdot W8(j) + a5 \cdot W16(j) + a6 \cdot M1(j) + a7 \cdot M2(j) + a8 \cdot M4(j) + a9 \cdot M8(j)$ (Eq. 10)

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As noted from Eq.10, an intended code symbol is obtained by multiplying each input information bit by the symbols of a corresponding basis Walsh code or basis mask sequence and summing the products.

In step 1640, sum indicating the achieved jth code symbol is output. j is increased by 1 in step 1650 and then the procedure returns to step 1610. Meanwhile, if j is 30 in step 1610, the encoding procedure ends.

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Now there will be given a description of the third embodiment of the decoding apparatus referring to FIG. 15. An input signal r(t) which includes the 30 encoded

symbols signal transmitted by a transmitter and two dummy symbols which have been inserted at the positions that have been punctured by the encoder is applied to 31 multipliers 1502 to 1506 and a correlation calculator 1520. A mask sequence generator 1500 generates all possible 31 mask sequences of length 32 M1 to M31. The multipliers 1502 to 1506 multiply the mask sequences received from the mask sequence generator 1500 by the input signal r(t). If a transmitter encoded TFCI bits with a predetermined mask sequence, one of the outputs of the multipliers 1502 to 1506 is free of the mask sequence, which means the mask sequence has no effect on the following correlation calculator. For example, if the transmitter used the mask sequence M31 for encoding the TFCI bits, the output of the multiplier 1506 that multiplies the mask sequence M31 by the input signal r(t) is free of the mask sequence. However, if the transmitter did not use a mask sequence, the input signal r(t) itself applied to a correlation calculator 1520 is a mask sequence-free signal. Each correlation calculators 1520 to 1526 calculates the correlation values of the outputs of the multipliers 1502 to 1506 with 64 bi-orthogonal codes of length 32, determines maximum correlation value among the 64-correlation sets, and outputs the determined maximum correlation values, the indexes of each biorthogonal codescorresponding to the determined maximum correlation values, and each indexe of the mask sequences to a correlation comparator 1540, respectively.

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The correlation comparator 1540 compares the 32 maximum correlation values received from the correlation calculators 1520 to 1526 and determines the largest of the maximum correlation values as a final maximum correlation. Then, the correlation comparator 1540 outputs the decoded TFCI bits transmitted by the transmitter on the basis of the indexes of the bi-orthogonal code and mask sequence corresponding to the final maximum correlation value. As in FIG. 17, the third embodiment of present invention can be also implemented by a single multiplier for multiplying the masks with r(t) and a single correlation calculator for calculating correlation values of bi-orthogonal codes.

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As described above, the present invention provides an apparatus and method for encoding and decoding a basic TFCI and an extended TFCI variably so that hardware is simplified. Another advantage is that support of both basic TFCI and extended TFCI error correcting coding schemes increases service stability. Furthermore, a minimum distance, a factor that determined the performance of an encoding apparatus, is large enough to satisfy the requirement of an IMT 2000 system, thereby ensuing excellent performance.

While the inverse preferred embodiments various changes in for spirit and scope of the in

in has been shown and described with reference to certain of, it will be understood by those skilled in the art that it details may be made therein without departing from the ion as defined by the appended claims.

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WHAT IS CLAIMED IS:

- 1. A transport format combination indicator (TFCI) encoding apparatus in a CDMA mobile communication system, comprising:
 - a one-bit generator for generating a sequence having the same symbols;
- a basis orthogonal sequence generator for generating a plurality of basis orthogonal sequences;
- a basis mask sequence generator for generating a plurality of basis mask sequences; and

an operation unit for receiving TFCI bits that are divided into a first information part representing biorthogonal sequence conversion, a second information part representing orthogonal sequence conversion, and a third information part representing mask sequence conversion and adding an orthogonal sequence selected from the basis orthogonal sequence based on the second information part and a mask sequence selected based on the third information part.

- 2. The TFCI encoding apparatus of claim 1, wherein the same symbols are 1s.
- 3. The TFCI encoding apparatus of claim 1, wherein the plurality of basis orthogonal sequences are a first Walsh code, a second Walsh code, a fourth Walsh code, an eighth Walsh code, and a sixteenth Walsh code.
- 4. The TFCI encoding apparatus of claim 1, wherein the basis mask sequences includes a first mask sequence "001010000110001111111000001110111", a second mask
- a first mask sequence "00101000011101111110000011101111", a second mask sequence "000000011100110110110111110001111", a fourth mask sequence "0000101011111001001101101010111", and an eighth mask sequence "0001110000110111001011111010001".
- 5. The TFCI encoding apparatus of claim 1, wherein the operation unit further comprises a converter for providing bi-orthogonal sequences by complementing the orthogonal sequences.
- 6. The TFCI encoding apparatus of claim 5, wherein the converter is an adder for adding a '1' to the symbols in each of the orthogonal sequences.

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- 7. The TFCI encoding apparatus of claim 1, wherein the basis mask sequence length is 32 symbols.
- 8. The TFCI encoding apparatus of claim 1, wherein the basis mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates a column transposition function to convert the sequences in the first group into the orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group into the mask sequences.
- 9. The TFCI encoding apparatus of claim 8, wherein the basis mask sequences are a first mask sequence "00101000011000111111000001110111", a second mask sequence "0000000111001101101101101101111001111", a fourth mask sequence "00001010111110010001101110010111", and an eighth mask sequence "000111000011011100101111101010001".
- 10. The TFCI encoding apparatus of claim 1, wherein the operation unit comprises:

a first multiplier for multiplying the same symbols by the first information part; a plurality of second multipliers for multiplying the basis orthogonal sequences by the respective TFCI bits representing the second information part;

a plurality of third multipliers for multiplying the basis mask sequences by the respective TFCI bits representing the third information part; and

an adder for adding the outputs of the first, second, and third multipliers.

11. A TFCI encoding apparatus in a CDMA mobile communication system, comprising:

an orthogonal sequence generator for generating a plurality of basis biorthogonal sequences;

a mask sequence generator for generating a plurality of basis mask sequences; and

an operation unit for adding a basis biorthogonal sequence and a basis mask sequence selected among the basis biorthogonal sequences and the basis mask sequences according to TFCI bits.

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- 12. The TFCI encoding apparatus of claim 11, wherein the plurality of basis biorthogonal sequences are a first Walsh code, a second Walsh code, a fourth Walsh code, an eighth Walsh code, a sixteenth Walsh code and an all "1" sequence which converts the orthogonal sequences to the biorthogonal sequences.
- 13. The TFCI encoding apparatus of claim 11, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to mask sequences.
- 14. The TFCI encoding apparatus of claim 11, wherein the basis mask sequences are a first mask sequence "001010000110001111110000011101111", a second mask sequence "00000001110011011011011011111000111", a fourth mask sequence "0000101011111001001101101010111", and an eighth mask sequence "0001110000110111100101111101010001".
- 15. The TFCI encoding apparatus of claim 11, wherein the operation unit comprises:
- a plurality of first multipliers for multiplying the basis biorthogonal sequences by corresponding TFCI bits;
- a plurality of second multipliers for multiplying the basis mask sequences by corresponding TFCI bits; and
- an adder for adding the outputs of the first and second multipliers and generating the sum as the TFCI sequence.
- 16. An apparatus for encoding TFCI bits including first information bits and second information bits in a CDMA mobile communication system, comprising:
- an orthogonal sequence generator for generating a plurality of biorthogonal sequences and outputting a biorthogonal sequence selected based on the first information bits among the plurality of biorthogonal sequences;

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a mask sequence generator for generating a plurality of mask sequences and outputting a mask sequence selected based on the second information bits among the plurality of mask sequences; and

an adder for adding the biorthogonal sequence and the mask sequence received from the orthogonal sequence generator.

- 17. The TFCI encoding apparatus of claim 12, wherein the plurality of biorthogonal sequences are Walsh codes and bi-orthogonal complement sequences of the Walsh codes.
- 18. The TFCI encoding apparatus of claim 16, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.
- 19. A TFCI encoding apparatus in a CDMA mobile communication system, comprising:

a one-bit generator for generating a sequence having the same symbols; an orthogonal sequence generator for generating a plurality of basis orthogonal sequences;

a mask sequence generator for generating a plurality of basis mask sequences;

a plurality of multipliers as many as input TFCI bits, for multiplying the same symbols by corresponding TFCI bits, the plurality of basis orthogonal sequences by corresponding TFCI bits, and the plurality of basis mask sequences by corresponding TFCI bits; and

an adder for summing sequences received from the plurality of multipliers.

- 20. The TFCI encoding apparatus of claim 15, wherein the same symbols are 1s.
 - 21. The TFCI encoding apparatus of claim 15, wherein the plurality of

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basis orthogonal sequences are a first Walsh code, a second Walsh code, a fourth Walsh code, an eighth Walsh code, and a sixteenth Walsh code.

- 22. The TFCI encoding apparatus of claim 19, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to the orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.
- 23. The TFCI encoding apparatus of claim 19, wherein the basis mask sequences are a first mask sequence "00101000011000111111000001110111", a second mask sequence "0000001110011011011011011111001111", a fourth mask sequence "0000101011111001001101101010111", and an eighth mask sequence "00011100001101110010111110101001".
- 24. A TFCI encoding method in a CDMA mobile communication system, comprising the steps of:

generating the same symbols;

generating a plurality of basis orthogonal sequences;

generating a plurality of basis mask sequences; and

receiving TFCI bits that are divided into a first information part representing biorthogonal sequence conversion, a second information part representing orthogonal sequence conversion, and a third information part representing mask sequence conversion and combining an orthogonal sequence selected from the basis orthogonal sequence based on the second information part, a biorthogonal sequence obtained by combining the selected orthogonal sequence with the same symbols selected based on the first information part, and a mask sequence selected based on the biorthogonal sequence and the third information part.

The TFCI encoding method of claim 24, wherein the same symbols are 1s.

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26. The TFCI encoding method of claim 24, wherein the plurality of basis orthogonal sequences are a first Walsh code, a second Walsh code, a fourth Walsh code, an eighth Walsh code, and a sixteenth Walsh code.

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27. The TFCI encoding apparatus of claim 24, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to the orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.

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28. The TFCI encoding method of claim 24, wherein the basis mask sequences are a first mask sequence "00101000011000111111000001110111", a second mask sequence "0000001110011011011011011111000111", a fourth mask sequence "0000101011111001000110110010111", and an eighth mask sequence "00011100001101110010111110100001".

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29. The TFCI encoding method of claim 24, wherein the same symbols are multiplied by the first information part, the basis orthogonal sequences are multiplied by the respective TFCI bits representing the second information part, the basis mask sequences are multiplied by the respective TFCI bits representing the third information part, and the multiplication results are summed.

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30. A TFCI encoding method in a CDMA mobile communication system, comprising the steps of:

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generating a plurality of basis biorthogonal sequences;
generating a plurality of basis mask sequences; and
adding a basis biorthogonal sequence and a basis mask sequence selected
among the basis biorthogonal sequences and the basis mask sequences according to

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TFCI bits.

31. The TFCI encoding method of claim 30, wherein the plurality of basis biorthogonal sequences are a first Walsh code, a second Walsh code, a fourth Walsh

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code, an eighth Walsh code, a sixteenth Walsh code and an all "1" sequence which converts the orthogonal sequences to the biorthogonal sequences.

32. The TFCI encoding apparatus of claim 30, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.

33. The TFCI encoding method of claim 30, wherein the basis mask sequences are a first mask sequence "001010000110001111110000011101111", a second mask sequence "00000001110011011011011011011111", a fourth mask sequence "00001010111110010001101101010111", and an eighth mask sequence "00011100001101110010111110101001".

- 34. The TFCI encoding method of claim 30, wherein the basis orthogonal sequences are multiplied by corresponding TFCI bits, the basis mask sequences are multiplied by corresponding TFCI bits, and the multiplication results are added to the TFCI sequence in the TFCI sequence generating step.
- 35. A method of encoding TFCI bits including first information bits and second information bits in a CDMA mobile communication system, comprising the steps of:

generating a plurality of biorthogonal sequences and outputting a biorthogonal sequence selected based on the first information bits among the plurality of biorthogonal sequences;

generating a plurality of mask sequences and outputting a mask sequence selected based on the second information bits among the plurality of mask sequences; and

adding the selected biorthogonal sequence and the selected mask sequence.

36. The TFCI encoding method of claim 35, wherein the plurality of

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biorthogonal sequences are Walsh codes and complement codes of the Walsh codes.

- 37. The TFCI encoding apparatus of claim 35, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.
- 38. A TFCI encoding method in a CDMA mobile communication system, comprising the steps of:

generating the same symbols;

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generating a plurality of basis orthogonal sequences;

generating a plurality of basis mask sequences;

receiving TFCI bits and multiplying the same symbols by corresponding TFCI bits, the plurality of basis orthogonal sequences by corresponding TFCI bits, and the plurality of basis mask sequences by corresponding TFCI bits; and

adding the multiplication results.

- 39. The TFCI encoding method of claim 38, wherein the same symbols are 1s.
- 40. The TFCI encoding method of claim 38, wherein the plurality of basis orthogonal sequences are a first Walsh code, a second Walsh code, a fourth Walsh code, an eighth Walsh code, and a sixteenth Walsh code.
- 41. The TFCI encoding apparatus of claim 38, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to the orthogonal sequences, inserts a column of '0' in the front of the sequences in the

second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.

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42. The TFCI encoding method of claim 38, wherein the basis mask sequences are a first mask sequence "00101000011000111111000001110111", a second mask sequence "0000000111001101101101101101111001111", and an eighth mask sequence "00011100001101111001011111010001".

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43. A TFCI decoding apparatus in a CDMA mobile communication system, comprising:

a mask sequence generator for generating at least one mask sequence;

at least one operation circuit for receiving an input signal and the generated mask sequence and removing the mask sequences from the input signal by multiplying the mask sequence by the input signal; and

at least one correlator for receiving the signal from the operation circuit, calculating correlation values of the received signal with a plurality of orthogonal sequences numbered with corresponding indexes, and selecting the largest of the calculated correlation value and the orthogonal sequence index corresponding to the largest correlation value.

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44. The TFCI encoding apparatus of claim 43, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.

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45. The TFCI decoding apparatus of claim 43, wherein the operation circuit is a multiplier.

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46. The TFCI decoding apparatus of claim 43, further comprising a

correlation comparator for determining the largest correlation value received from a plurality of correlators and generating an orthogonal sequence index and a mask sequence index corresponding to the largest correlation value.

- 47. The TFCI decoding apparatus of claim 46, wherein the mask sequence index is the index of the mask sequence used to remove a mask sequence from the input signal.
- 48. A TFCI decoding apparatus in a CDMA mobile communication system, comprising;
- a mask sequence generator for sequentially generating a plurality of mask sequences;

an operation circuit for receiving an input signal and the mask sequences from the mask sequence generator, and removing a mask sequence form the input signal by multiplying the mask sequences by the input signal;

a correlator for receiving signals from the operation circuit sequentially, calculating correlation value of each received signal with a plurality of orthogonal sequences having corresponding indexes, and sequentially selecting the largest correlation values and an orthogonal sequence index corresponding to the largest correlation value; and

a correlation comparator for determining the highest correlation value out of the sequentially selected largest correlation values, from the correlator and outputting an orthogonal sequence index and a mask sequence index corresponding to the determined highest correlation value.

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- 49. The TFCI encoding apparatus of claim 48, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.
 - 50. The TFCI decoding apparatus of claim 48, further comprising a

memory for storing the input signal and outputting the input signal to the operation circuit until the input signal is completely multiplied by the mask sequences generated from the mask sequence generator.

51. The TFCI decoding apparatus of claim 50, wherein the operation circuit is a multiplier.

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- 52. The TFCI decoding apparatus of claim 48, wherein the mask sequence index is the index of the mask sequence used to remove a mask sequence from the input signal.
- 53. A TFCI decoding apparatus in a CDMA mobile communication system, comprising;
- a mask sequence generator for sequentially generating a plurality of mask sequences;
- a plurality of operation circuits for receiving an input signal and the mask sequences from the mask sequence generator and multiplying the mask sequences by the input signal;
- a first correlator for calculating correlation values of the received signal with a plurality of orthogonal sequences, selecting the largest correlation value and an orthogonal sequence index corresponding to the largest correlation value;
- a plurality of secondary correlators for receiving the input signal and the outputs of the operation circuits, calculating correlation values of the received signals with a plurality of orthogonal sequences having corresponding indexes, and selecting the largest correlation value and orthogonal sequences index corresponding to the largest correlation value, respectably; and
- a correlation comparator for determining the highest correlation value from the selected largest correlation values received from the correlators and outputting TFCI information based on an orthogonal sequence index and a mask sequence index corresponding to the determined highest correlation value.
- 54. The TFCI encoding apparatus of claim 53, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to

orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.

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- 55. The TFCI decoding apparatus of claim 54, wherein the operation circuits are multipliers.
- 56. The TFCI decoding apparatus of claim 53, wherein the mask sequence index is the index of the mask sequence used to remove a mask sequence from the input signal corresponding to the determined correlation value.
 - 57. A TFCI decoding method in a CDMA mobile communication system, comprising the steps of:

generating at least one mask sequence;

receiving an input signal and the mask sequence and removing a mask sequence from the input signal by multiplying the mask sequence by the input signal;

receiving the product signal, calculating correlation values of the product signal with a plurality of orthogonal sequences having corresponding indexes; and

selecting the largest correlation value from the calculated correlation values and outputting an orthogonal sequence index corresponding to the largest correlation value.

- 58. The TFCI encoding apparatus of claim 57, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to convert the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.
- 59. The TFCI decoding method of claim 57, further comprising the step of determining the highest correlation value from the selected largest correlation values obtained by selecting the largest correlation value from the calculated correlation values;

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and outputting an orthogonal sequence index and a mask sequence index corresponding to the determined highest correlation value.

- 60. The TFCI decoding apparatus of claim 59, wherein the mask sequence index is the index of the mask sequence used to remove a mask sequence from the input signal corresponding to the highest correlation value.
- 61. A TFCI decoding method in a CDMA mobile communication system, comprising the steps of:

generating a plurality of mask sequences;

receiving an input signal and the mask sequences and removing a mask sequence from the input signal by multiplying the mask sequences by the input signal;

receiving the product signals, calculating correlation values of each of the product signals with a plurality of orthogonal sequences having corresponding indexes, and selecting the largest correlation values and orthogonal sequence indexes corresponding to the largest correlation values; and

determining the highest correlation value from the largest correlation values and outputting an orthogonal sequence index and a mask sequence index corresponding to the determined highest correlation value.

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62. The TFCI encoding apparatus of claim 61, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.

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63. The TFCI decoding method of claim 61, wherein the mask sequence index is the index of the mask sequence used to remove a mask sequence from the input signal corresponding to the highest correlation value.

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64. A TFCI decoding method in a CDMA mobile communication system, comprising the steps of:

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generating a plurality of mask sequences;

receiving an input signal and the mask sequences and multiplying each mask sequence by the input signal;

receiving the multiplied signals and calculating correlation values of each of the received multiplied signals with a plurality of orthogonal sequences having corresponding indexes;

selecting the largest correlation value among the calculated correlation values for each of the multiplied signals and an orthogonal sequence index corresponding to the largest correlation value; and ;

determining the highest correlation value from all of the largest correlation values and an orthogonal code index corresponding to the highest correlation value

- 65. The TFCI encoding apparatus of claim 64, wherein the mask sequence generator has a first m-sequence and a second m-sequence which can be added together to form a Gold code, forms a first sequence group having sequences formed by cyclically shifting the first m-sequence and a second sequence group having sequences formed by cyclically shifting the second m-sequence, generates and applies a column transposition function to the sequences in the first group to orthogonal sequences, inserts a column of '0' in the front of the sequences in the second group, and generates and applies a reverse column transposition function to the sequences in the second group to convert the sequences in the second group to the mask sequences.
- 66. The TFCI decoding method of claim 64, wherein the mask sequence index is the index of the mask sequence used to remove a mask sequence from the input signal corresponding to the highest correlation value.
- 67. A mask sequence generating method for use in a TFCI encoding and decoding, comprising the steps of:

selecting a Gold sequence which is determined by adding a first m-sequence and a second m-sequence, each of the m-sequences generated by different generation polynomials;

generating a first m-sequence group by cyclically shifting the first m-sequence where the first m-sequence is shifted one to 'n'times, 'n' is a length of the first and second m-sequences and each shift of the first m-sequence produces a sequence forming the first m-sequence group;

generating a second m-sequence group by cyclically shifting the second m-

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sequence where the second m-sequence is shifted one to 'n' times and each shift of the second m-sequence produces a sequence forming the second m-sequence group;

determining a column transposition function that converts sequences in the first m-sequence group to orthogonal sequences;

inserting a column of '0' in the front of the sequences in the second m-sequence group;

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column changing the second m-sequence group by applying the reverse function of the sequence transposition function to generate mask sequences of the TFCI coding/decoding.

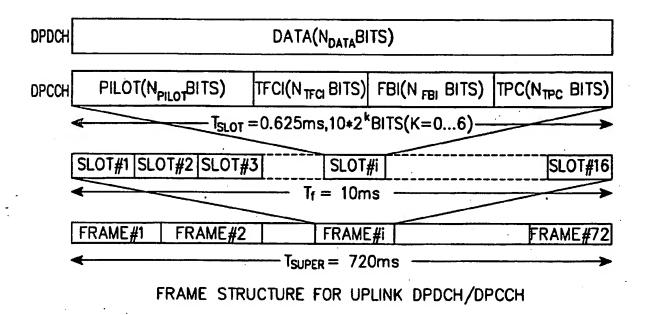


FIG. 1A

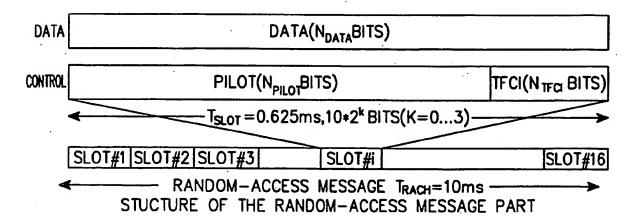
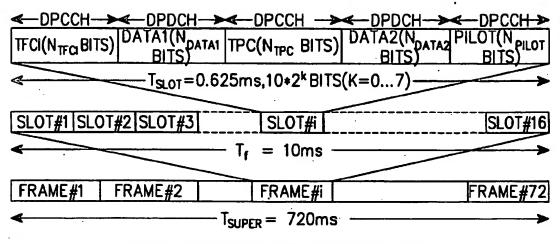


FIG. 1B



FRAME STRUCTURE OF DOWNLINK DPCH

FIG. 1C

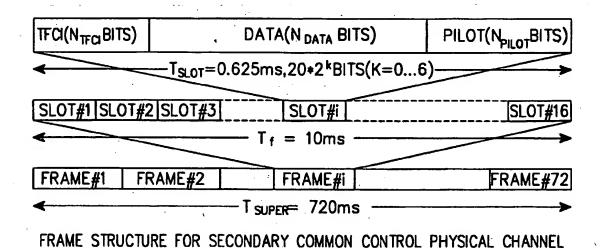
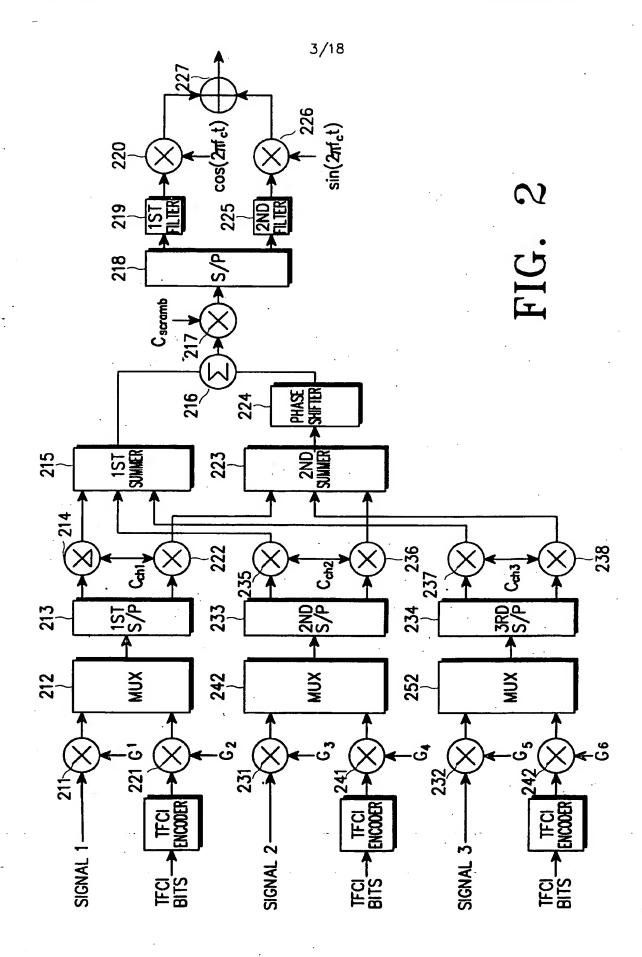


FIG. 1D

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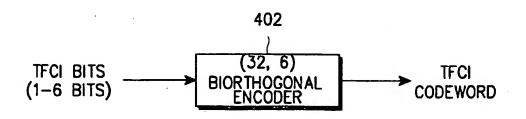


FIG. 4A

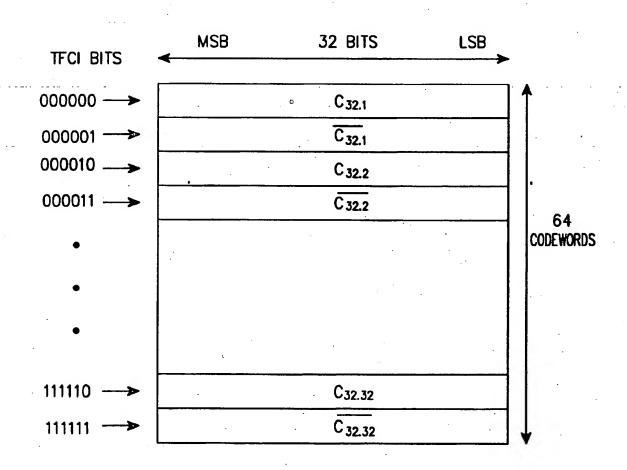


FIG. 4B

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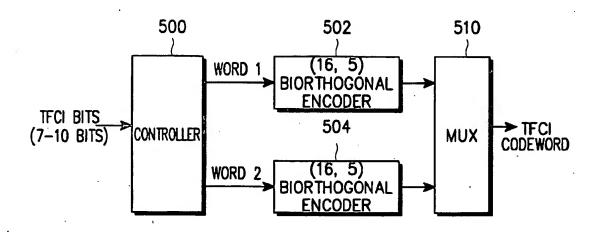


FIG. 5A

TFCI =
$$a_{10}a_{9}a_{8}a_{7}a_{6}a_{5}a_{4}a_{3}a_{2}a_{1}$$

 $n = (MAXIMUM INTEGER EQUAL TO OR SMALLER THAN(TFCI)$
IF TFCI < $n^{2} + n$
THEN WORD1 = n ; WORD2 = TFCI- n^{2}
ELSE WORD1 = TFCI - n^{2} ; WORD2 = n

FIG. 5B

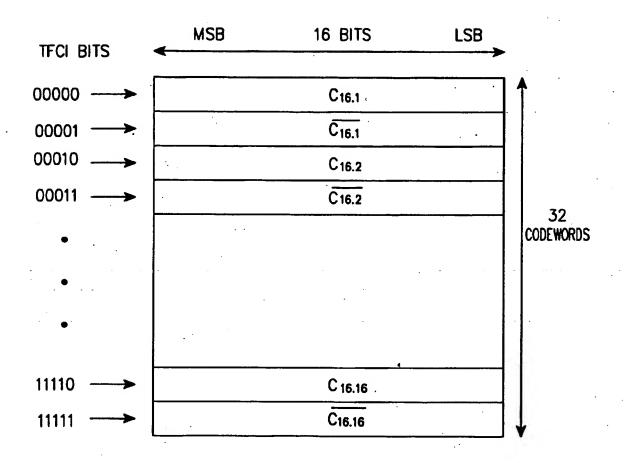
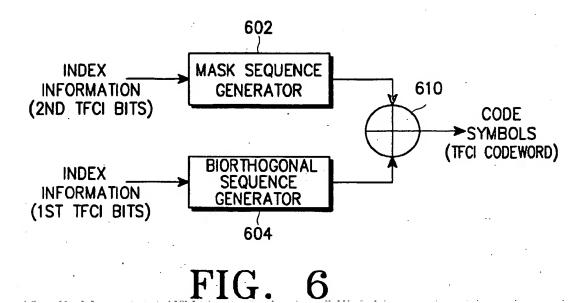


FIG. 5C

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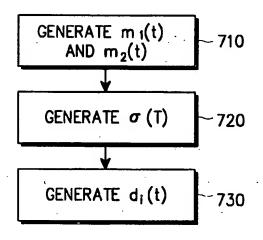
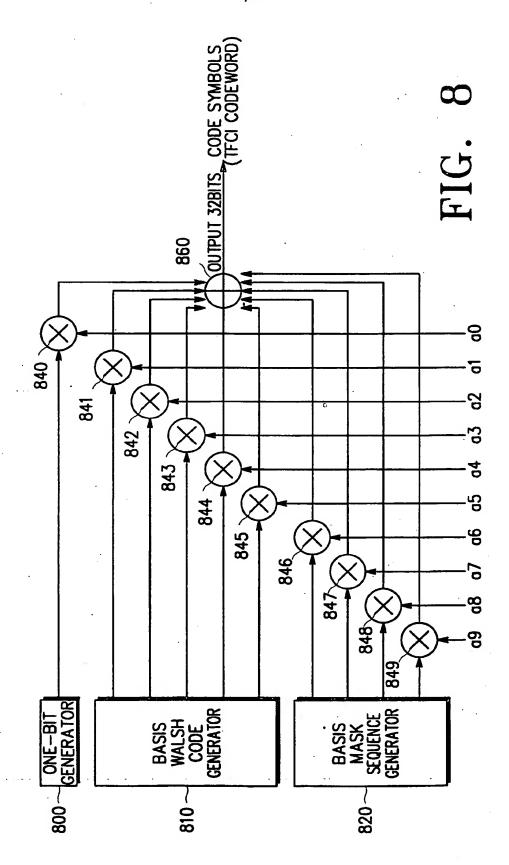


FIG. 7



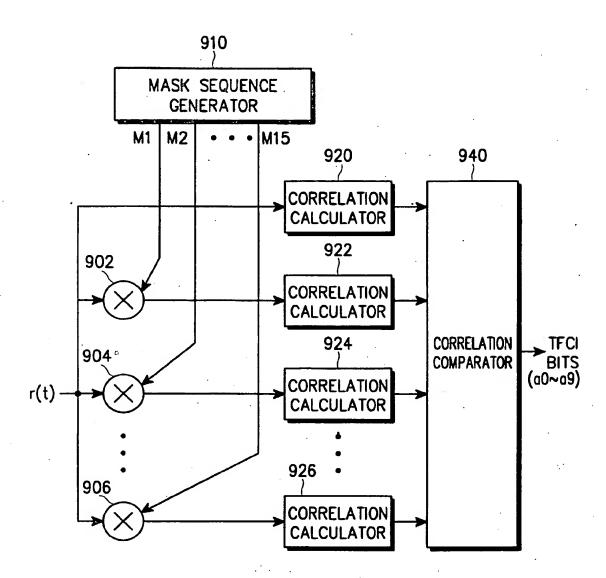


FIG. 9

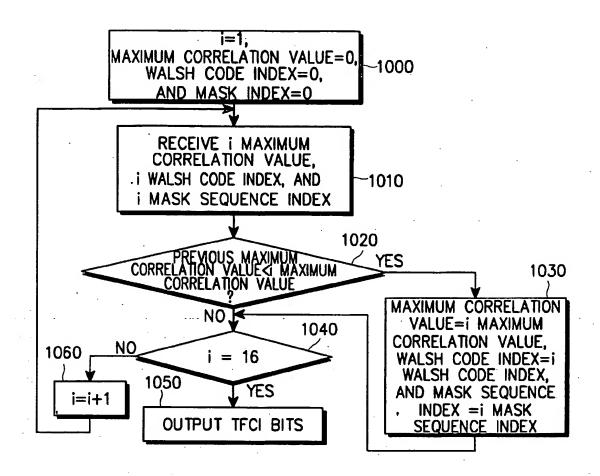


FIG. 10

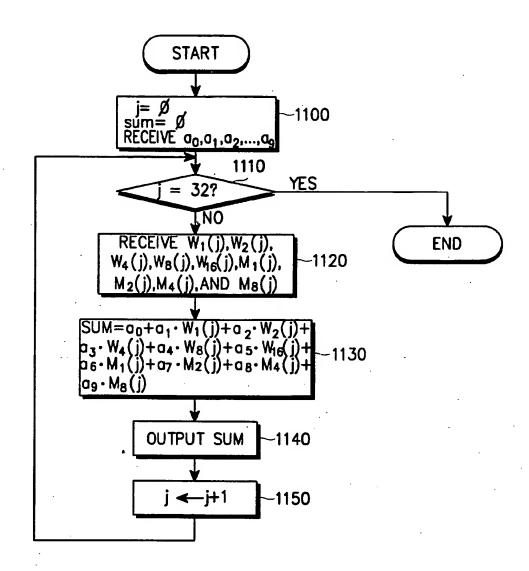


FIG. 11

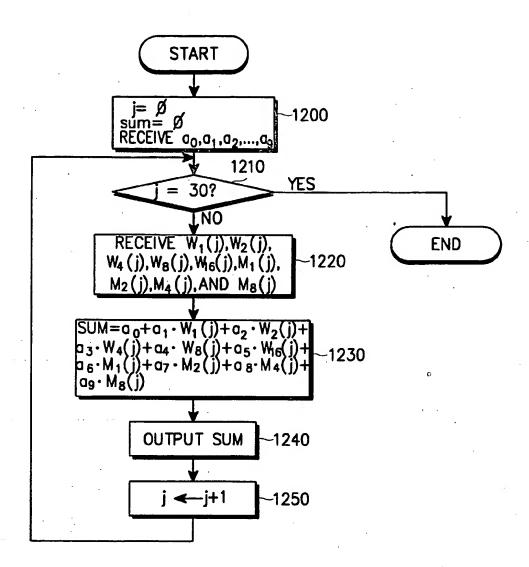


FIG. 12

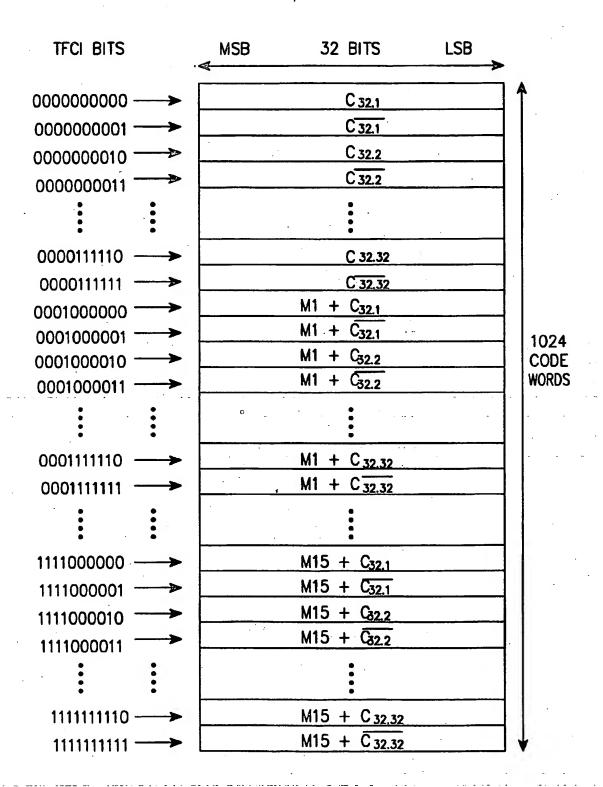
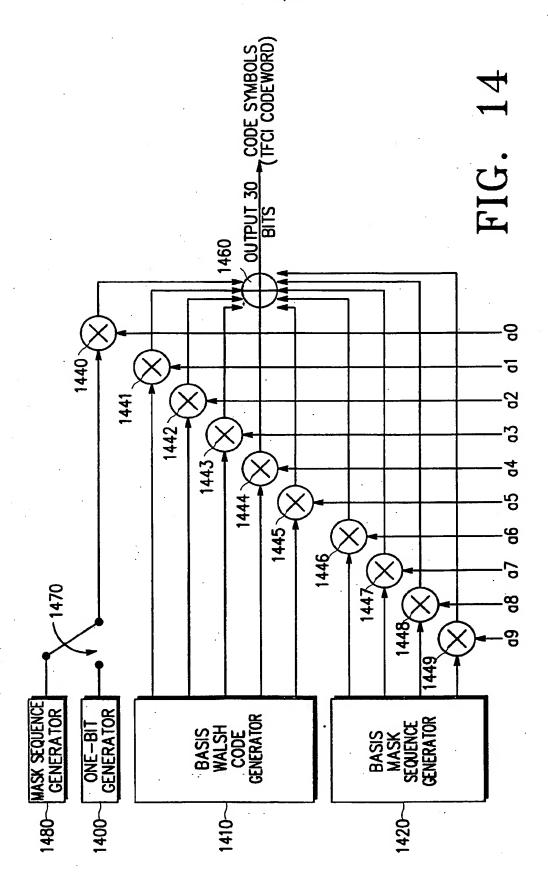


FIG. 13



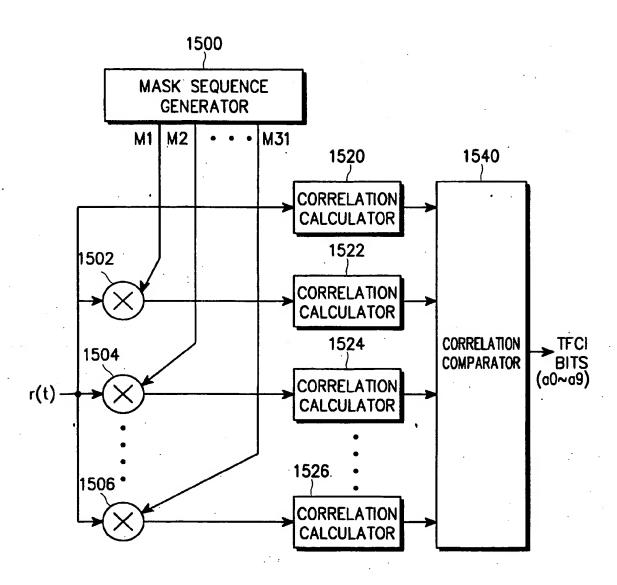
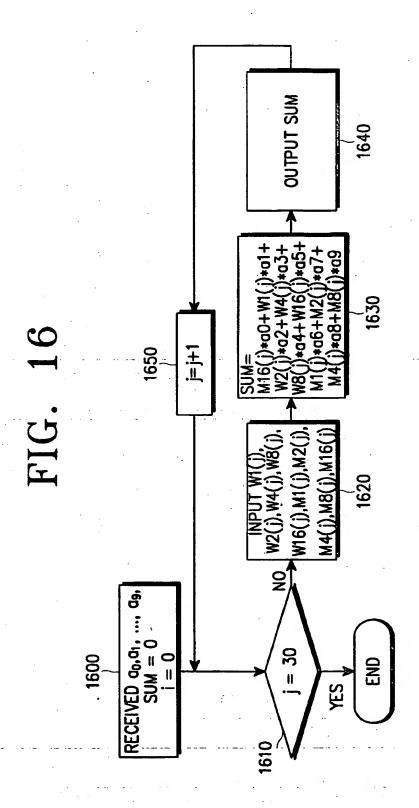
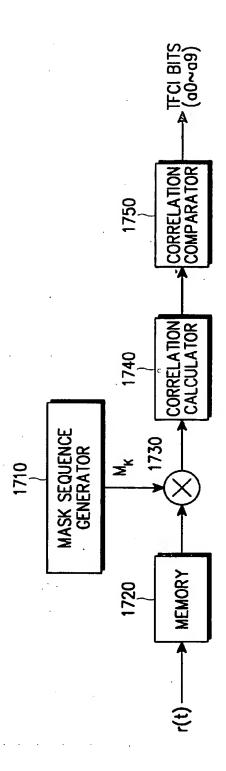


FIG. 15







International application No. PCT/KR00/00731

CLASSIFICATION OF SUBJECT MATTER

IPC7 H04L 9/06

According to International Patent Classification (IPC) or to both national classification and IPC

FIELDS SEARCHED

Minimun documentation searched (classification system followed by classification symbols) IPC7 H04L 1/00, 9/00,12/00 H04J

Documentation searched other than minimun documentation to the extent that such documents are included in the fileds searched KOREAN PATENTS AND APPLICATIONS FOR INVENTIONS SINCE 1983 JAPANESE PATENTS AND APPLICATIONS FOR INVENTIONS

Electronic data base consulted during the intertnational search (name of data base and, where practicable, search trerms used) WPI

DOCUMENTS CONSIDERED TO BE RELEVANT

Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	KR 99-15261 A (ETRI) 05 MARCH 1999(05.03.1999) abstract, pgae 3, fig1	1-3
P.Y	KR 99-75942 A (DAEWOO ELECTRONICS CORP.) 15 OCTOBER 1999 (15.10.1999) abstract, page3 lines 6 to 21	1-3
P.Y	KR 99-76303 A (DAEWOO ELECTRONICS CORP.) 15 OCTOBER 1999(15.10.1999) abstract, page 3 lines 23 to 39	1-3
P. A	KR 2000-31698 A(L.G ELECTRONICS CORP.) 05 JUNE 2000(05.06.2000) figl, page3 lines 15 to 34	1-3

	figl, page3 lines 15 to 34		
X Furthe	r documents are listed in the continuation of Box C.	X See patent family annex.	
"A" document to be of pa "E" earlier app filing date "L" document cited to es special rea "O" document means "P" document	ategories of cited documents: defining the general state of the art which is not considered urticular relevence plication or patent but published on or after the international which may throw doubts on priority claim(s) or which is stablish the publication date of citation or other ason (as specified) referring to an oral disclosure, use, exhibition or other published prior to the international filing date but later riority date claimed	"T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention "X" document of particular relevence; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone "Y" document of particular relevence; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art "&" document member of the same patent family	
Date of the actual completion of the international search 22 SEPTEMBER 2000 (22.09.2000)		Date of mailing of the international search report 25 SEPTEMBER 2000 (25.09.2000)	
Korean Indu Government	ailing address of the ISA/KR strial Property Office Complex-Taejon, Dunsan-dong, So-ku, Taejon n City 302-701, Republic of Korea	Authorized officer LEE, Son Tack	

Telephone No. 82-42-481-5667

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Facsimile No. 82-42-472-7140

International application No.
PCT/KR00/00731

C (Continuat	C (Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT				
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.			
A	EP 565506 A(ERICSSON, GE MOBILE COMMUNICATIONS INC) 01 APRIL 1993(01.04.1993) abstract	1-3			
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International application No.
PCT/KR00/00731

Box I Observations where cert	tain claims were found unsearchable (Continuation of item 1 of first sheet)			
This international search report has not been established in respect of certain claims under Article 17(2)(a) for the following reasons:				
1. Claims Nos.: because they relate to subject	ct matter not required to be searched by this Authority, namely:			
2. Claims Nos.: because they relate to part of extent that no meaningful in	of the international application that do not comply with the prescribed requirements to such an ternational search can be carried out, specifically:			
3. X Claims Nos.:27,32,37,etc. because they are dependent	claims and are not drafted in accordance with the second and third sentences of Rule 6.4(a).			
Box II Observations where un	ity of invention is lacking (Continuation of item 2 of first sheet)			
This International Search Authority fo	und multiple inventions in this international application, as follows:			
	•			
·				
As all required additional sea claims.	arch fees were timely paid by the applicant, this international search report covers all searchable			
	ald be established without effort justifying an additional fee, this Authority did not invite payment			
of any addition fee.				
3. As only some of the required additional search fees were timely paid by the applicant, this international search report covers only those claims for which fees were paid, specifically claims Nos.:				
	•			
*				
4. No required additional search fees were timely paid by the applicant. Consequently, this international search report is restricted to the invention first mentioned in the claims; it is covered by claims Nos.:				
	· · · · · · · · · · · · · · · · · · ·			
Remark on Protest	The additional search fees were accompanied by the applicant's protest.			
	No protest accompanied the payment of additional search fees.			

Information on patent family members

International application No.
PCT/KR00/00731

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
KR 99-15261 A	05.03.1999	NONE	
KR 99-75942 A	15.10.1999	NONE	·
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